A 13.3 mΩ, 4 A Integrated Power Switch with 12 V / 24 V V_{IN} Lockout Select and MOSFET Current Monitor Output

General Description

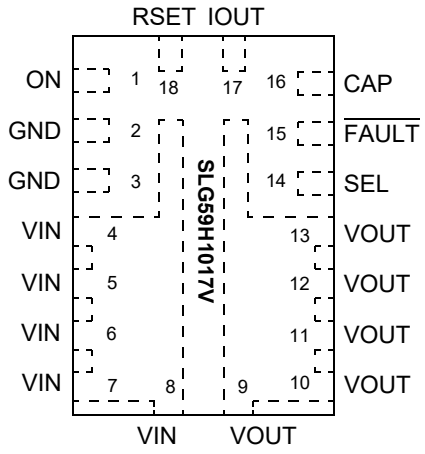
The SLG59H1017V is a high-performance 13.3 mΩ NMOS power switch designed to control 12 V or 24 V power rails up to 4 A. Using a proprietary MOSFET design, the SLG59H1017V achieves a stable 13.3 mΩ RDS_{ON} across a wide input voltage range. In combining novel FET design and copper pillar interconnects, the SLG59H1017V package also exhibits a low thermal resistance for high-current operation.

Designed to operate over a -40 °C to 85 °C range, the SLG59H1017V is available in a low thermal resistance, RoHS-compliant, 1.6 x 3.0 mm STQFN package.

Features

- Wide Operating Input Voltage: 12 V or 24 V
- Maximum Continuous Current: 4 A
- Automatic nFET SOA Protection
- High-performance MOSFET Switch
  - Low RDS_{ON}: 13.3 mΩ at V_{IN} = 24 V
  - Low $\Delta RDS_{ON}/\Delta V_{IN}$: < 0.05 mΩ/V
  - Low $\Delta RDS_{ON}/\Delta T$: < 0.06 mΩ/°C
- Pin-selectable 12V/24V Input Overvoltage and Undervoltage Lockout
- Capacitor-adjustable Inrush Current Control
- Two stage Current Limit Protection:
  - Resistor-adjustable Active Current Limit
  - Internal Short-circuit Current limit
- Open Drain $\overline{FAULT}$ Signaling
- Analog MOSFET Current Monitor Output: 10 µA/A
- Fast 4 kΩ Output Discharge
- Pb-Free / Halogen-Free / RoHS Compliant Packaging

Pin Configuration

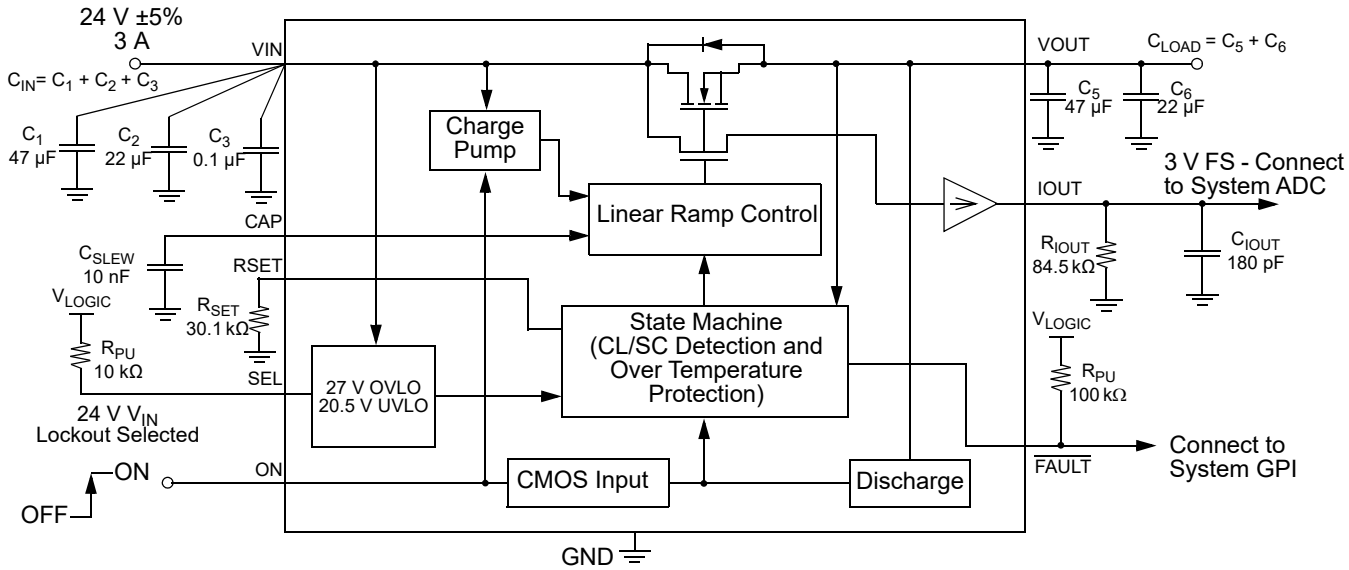


18-pin STQFN
1.6 x 3.0 mm, 0.40mm pitch
(Top View)

Applications

- Power-Rail Switching
- Multifunction Printers
- Large-format Copiers
- Telecommunications Equipment
- High-performance Computing
  - 12 V and 24 V Point-of-Load Power Distribution
- Motor Drives

Block Diagram and 3 A Typical Application Circuit



Pin Description

Pin #	Pin Name	Type	Pin Description
1	ON	Input	A low-to-high transition on this pin initiates the operation of the SLG59H1017V's state machine. ON is an asserted HIGH, level-sensitive CMOS input with ON_V_{IL} < 0.3 V and ON_V_{IH} > 0.9 V. As the ON pin input circuit does not have an internal pull-down resistor, connect this pin to a general-purpose output (GPO) of a microcontroller, an application processor, or a system controller, do not allow this pin to be open-circuited.
2	GND	GND	Pin 2 is a low-current GND terminal for the SLG59H1017V. Connect directly to Pin 3.
3	GND	GND	Pin 3 is the main ground connection for the SLG59H1017V's internal charge pump, its gate driver and current-limit circuits as well as its internal state machine. Therefore, use a short, stout connection from Pin 3 to the system's analog or power plane.
4-8	VIN	MOSFET	VIN supplies the power for the operation of the SLG59H1017V, its internal control circuitry, and the drain terminal of the nFET power switch. With 5 pins fused together at VIN, connect a 47 µF (or larger) low-ESR capacitor from this pin to ground. Capacitors used at VIN should be rated at 50 V or higher.
9-13	VOUT	MOSFET	Source terminal of n-channel MOSFET (5 pins fused for VOUT). Connect a 47 µF (or larger) low-ESR capacitor from this pin to ground. Capacitors used at VOUT should be rated at 50 V or higher.
14	SEL	Input	As a low logic-level CMOS input with SEL_V_{IL} < 0.3 V and SEL_V_{IH} > 1.65 V, SEL selects one of two undervoltage/overvoltage lockout windows. When SEL = LOW, the V_{IN} undervoltage/overvoltage lockout window is set for 12 V ±10% applications. When SEL = HIGH, the V_{IN} undervoltage/overvoltage lockout window is set for 24 V ± 5% applications. See the Electrical Characteristics table for additional information.
15	$\overline{FAULT}$	Output	An open drain output, $\overline{FAULT}$ is asserted within $T\overline{FAULT}_{LOW}$ when a V_{IN} undervoltage, V_{IN} overvoltage, a current-limit, or an over-temperature condition is detected. $\overline{FAULT}$ is deasserted within $T\overline{FAULT}_{HIGH}$ when the fault condition is removed. Connect an 100 kΩ external resistor from the $\overline{FAULT}$ pin to local system logic supply.
16	CAP	Output	A low-ESR, stable dielectric, ceramic surface-mount capacitor connected from CAP pin to GND sets the V_{OUT} slew rate and overall turn-on time of the SLG59H1017V. For best performance, the range for C_{SLEW} values are 10 nF ≤ C_{SLEW} ≤ 20 nF – please see typical characteristics for additional information. Capacitors used at the CAP pin should be rated at 10 V or higher. Please consult Applications Section on how to select C_{SLEW} based on V_{OUT} slew rate and loading conditions.
17	IOUT	Output	IOUT is the SLG59H1017V's power MOSFET load current monitor output. As an analog current output, this signal when applied to a ground-reference resistor generates a voltage proportional to the current through the n-channel MOSFET. The I_{OUT} transfer characteristic is typically 10 µA/A with a voltage compliance range of 0.5 V ≤ V_{IOUT} ≤ 4 V. Optimal I_{OUT} linearity is exhibited for 0.5 A ≤ I_{DS} ≤ 4 A. In addition, it is recommended to bypass the IOUT pin to GND with a 0.18 nF capacitor.
18	RSET	Input	A 1%-tolerance, metal-film resistor between 20 kΩ and 91 kΩ sets the SLG59H1017V's active current limit. A 91 kΩ resistor sets the SLG59H1017V's active current limit to 1 A and a 20 kΩ resistor sets the active current limit to 4.5 A.

Ordering Information

Part Number	Type	Production Flow
SLG59H1017V	STQFN 18L FC	Industrial, -40 °C to 85 °C
SLG59H1017VTR	STQFN 18L FC (Tape and Reel)	Industrial, -40 °C to 85 °C

Absolute Maximum Ratings

Parameter	Description	Conditions	Min.	Typ.	Max.	Unit
V_{IN} to GND	Power Switch Input Voltage to GND	Continuous	-0.3	--	30	V
		Maximum pulsed V_{IN}, pulse width <0.1s	--	--	32	V
V_{OUT} to GND	Power Switch Output Voltage to GND		-0.3	--	V_{IN}	V
ON, SEL, CAP, RSET, IOUT, and FAULT to GND	ON, SEL, CAP, RSET, IOUT, and FAULT Pin Voltages to GND		-0.3	--	7	V
T_S	Storage Temperature		-65	--	150	°C
ESD_{HBM}	ESD Protection	Human Body Model	2000	--	--	V
ESD_{CDM}	ESD Protection	Charged Device Model	500	--	--	V
MSL	Moisture Sensitivity Level			1		
θ_{JA}	Package Thermal Resistance, Junction-to-Ambient	1.6 x 3.0 mm 18L STQFN; Determined with the device mounted onto a 1 in^2, 1 oz. copper pad of FR-4 material	--	40	--	°C/W
MOSFET IDS_{CONT}	Continuous Current from VIN to VOUT	T_J < 150 °C	--	--	4	A
MOSFET IDS_{PEAK}	Peak Current from VIN to VOUT	Maximum pulsed switch current, pulse width < 1 ms	--	--	6	A

Note: Stresses greater than those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

Electrical Characteristics

12 V ≤ V_{IN} ≤ 24 V; C_{IN} = 47 µF, T_A = -40 °C to 85 °C, unless otherwise noted. Typical values are at T_A = 25 °C

Parameter	Description	Conditions	Min.	Typ.	Max.	Unit
V_{IN}	Operating Input Voltage		10.8	--	25.2	V
$V_{IN(OVLO)}$	V_{IN} Overvoltage Lockout Threshold	V_{IN} ↑; SEL = HIGH	25.3	27	28.5	V
		V_{IN} ↑; SEL = LOW	13.3	13.7	14.5	V
$V_{IN(UVLO)}$	V_{IN} Undervoltage Lockout Threshold	V_{IN} ↓; SEL = HIGH	19.5	20.5	21.5	V
		V_{IN} ↓; SEL = LOW	9.7	10.2	10.7	V
I_Q	Quiescent Supply Current	ON = HIGH; I_{DS} = 0 A	--	0.5	0.6	mA
I_{SHDN}	OFF Mode Supply Current	ON = LOW; I_{DS} = 0 A	--	1	3	µA
RDS_{ON}	ON Resistance	T_A = 25 °C; I_{DS} = 0.1 A	--	13.3	14	mΩ
		T_A = 85 °C; I_{DS} = 0.1 A	--	16.8	19	mΩ
MOSFET IDS	Current from VIN to VOUT	Continuous	--	--	4	A
I_{LIMIT}	Active Current Limit, I_{ACL}	V_{OUT} > 0.5 V; R_{SET} = 30.1 kΩ	3.0	3.19	3.5	A
	Short-circuit Current Limit, I_{SCL}	V_{OUT} < 0.5 V	--	0.5	-	A
T_{ACL}	Active Current Limit Response Time	V_{IN} = 12 V; R_{SET} = 91 kΩ; C_{LOAD} = 10 µF; switch in 10 Ω load	117	150	220	µs
$R_{DISCHRG}$	Output Discharge Resistance	V_{OUT} = 0.4 V Input Bias; ON = LOW	3.5	4.4	5.3	kΩ

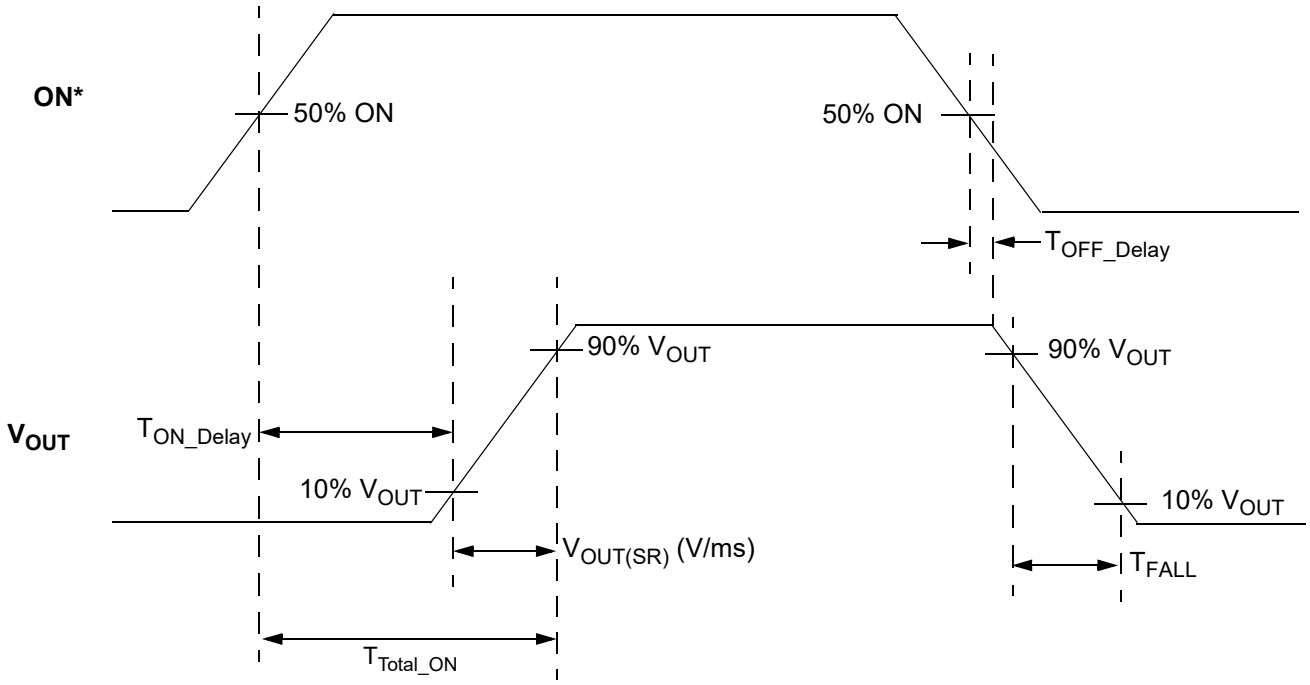
Electrical Characteristics (continued)

12 V ≤ V_{IN} ≤ 24 V; C_{IN} = 47 µF, T_A = -40 °C to 85 °C, unless otherwise noted. Typical values are at T_A = 25 °C

Parameter	Description	Conditions	Min.	Typ.	Max.	Unit
I_{OUT}	Analog MOSFET Current Monitor Output	I_{DS} = 1 A	9.3	10	10.7	µA
		I_{DS} = 3 A	28.5	30	31.5	µA
T_{IOUT}	I_{OUT} Response Time to Change Current in Main MOSFET	C_{IOUT} = 180 pF; Step load 0 to 2.4 A; 0% to 90% I_{OUT}	--	45	--	µs
C_{LOAD}	Output Load Capacitance	C_{LOAD} connected from VOUT to GND	--	47	--	µF
T_{ON_Delay}	ON Delay Time	50% ON to 10% V_{OUT} ↑; V_{IN} = 12 V; C_{SLEW} = 10 nF; R_{LOAD} = 100 Ω, C_{LOAD} = 10 µF	480	600	720	µs
		50% ON to 10% V_{OUT} ↑; V_{IN} = 24 V; C_{SLEW} = 10 nF; R_{LOAD} = 100 Ω, C_{LOAD} = 10µF	0.8	1.0	1.2	ms
T_{Total_ON}	Total Turn ON Time	50% ON to 90% V_{OUT} ↑	Set by External C_{SLEW}[1]			ms
		50% ON to 90% V_{OUT} ↑; V_{IN} = 12 V; C_{SLEW} = 10 nF; R_{LOAD} = 100 Ω, C_{LOAD} = 10 µF	2.9	3.6	4.3	ms
		50% ON to 90% V_{OUT} ↑; V_{IN} = 24 V; C_{SLEW} = 10 nF; R_{LOAD} = 100 Ω, C_{LOAD} = 10 µF	5.7	7.1	8.5	ms
$V_{OUT(SR)}$	V_{OUT} Slew rate	10% V_{OUT} to 90% V_{OUT} ↑	Set by External C_{SLEW}[1]			V/ms
		10% V_{OUT} to 90% V_{OUT} ↑; V_{IN} = 12 V or 24 V; C_{SLEW} = 10 nF; R_{LOAD} = 100 Ω, C_{LOAD} = 10 µF	2.7	3.2	3.9	V/ms
T_{OFF_Delay}	OFF Delay Time	50% ON to V_{OUT} Fall Start ↓; V_{IN} = 12 V or 24 V; R_{LOAD} = 100 Ω, No C_{LOAD}	--	15	--	µs
T_{FALL}	V_{OUT} Fall Time	90% V_{OUT} to 10% V_{OUT}; ON = HIGH-to-LOW; V_{IN} = 12 V or 24 V; R_{LOAD} = 100 Ω, No C_{LOAD}	10.4	12.7	14.3	µs
$TFAULT_{LOW}$	$\overline{FAULT}$ Assertion Time	Abnormal Step Load Current event to $\overline{FAULT}$↓; I_{ACL} = 1 A; V_{IN} = 24 V; R_{SET} = 91 kΩ; switch in 20 Ω load	--	80	--	µs
$TFAULT_{HIGH}$	$\overline{FAULT}$ De-assertion Time	Delay to $\overline{FAULT}$↑ after fault condition is removed; I_{ACL} = 1 A; V_{IN} = 24 V; R_{SET} = 91 kΩ; switch out 20 Ω load	--	180	--	µs
$\overline{FAULT}_{VOL}$	$\overline{FAULT}$ Output Low Voltage	$I_{\overline{FAULT}}$ = 1 mA	--	0.2	--	V
ON_V_{IH}	ON Pin Input High Voltage		0.9	--	5	V
ON_V_{IL}	ON Pin Input Low Voltage		-0.3	0	0.3	V
SEL_V_{IH}	SEL pin Input High Voltage		1.65	--	4.5	V
SEL_V_{IL}	SEL pin Input Low Voltage		-0.3	--	0.3	V
$I_{ON(Leakage)}$	ON Pin Leakage Current	1 V ≤ ON ≤ 5 V or ON = GND	--	--	1	µA
$THERM_{ON}$	Thermal Protection Shutdown Threshold		--	145	--	°C
$THERM_{OFF}$	Thermal Protection Restart Threshold		--	125	--	°C

Notes:
1. Refer to typical Timing Parameter vs. C_{SLEW} performance charts for additional information.

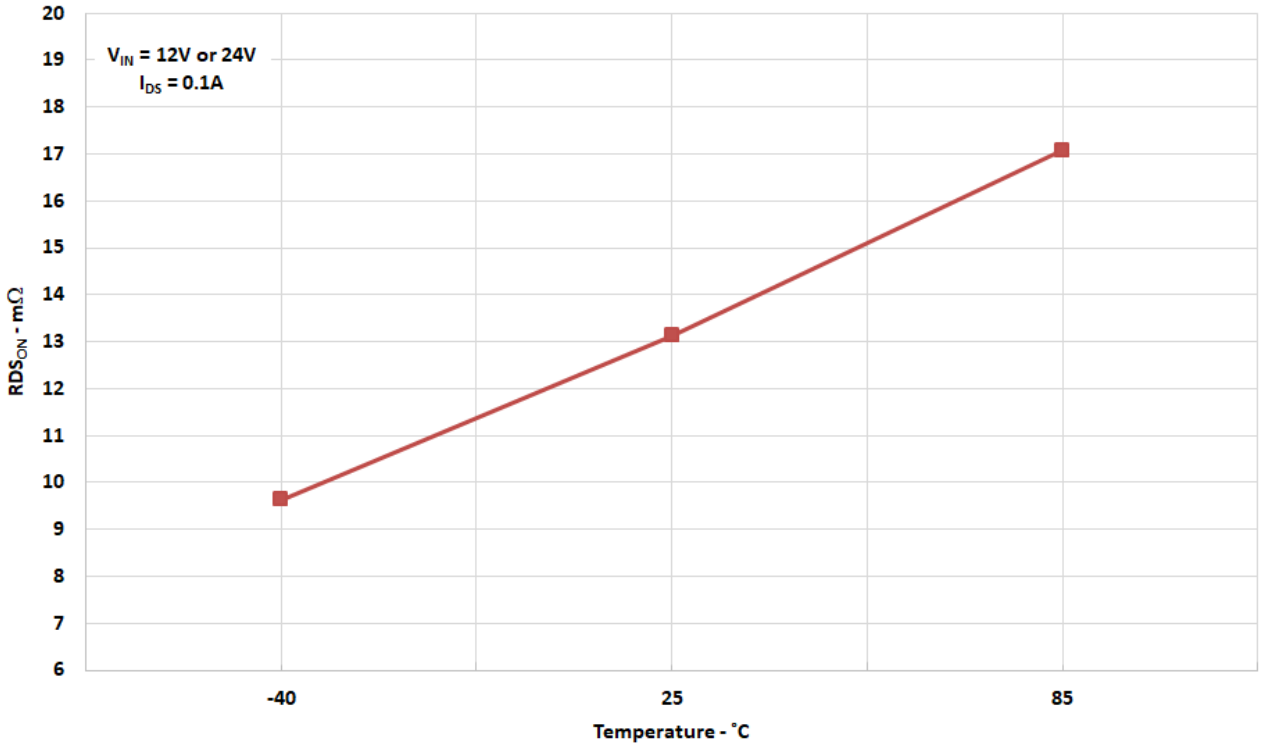
T_{Total_ON}, T_{ON_Delay} and Slew Rate Measurement



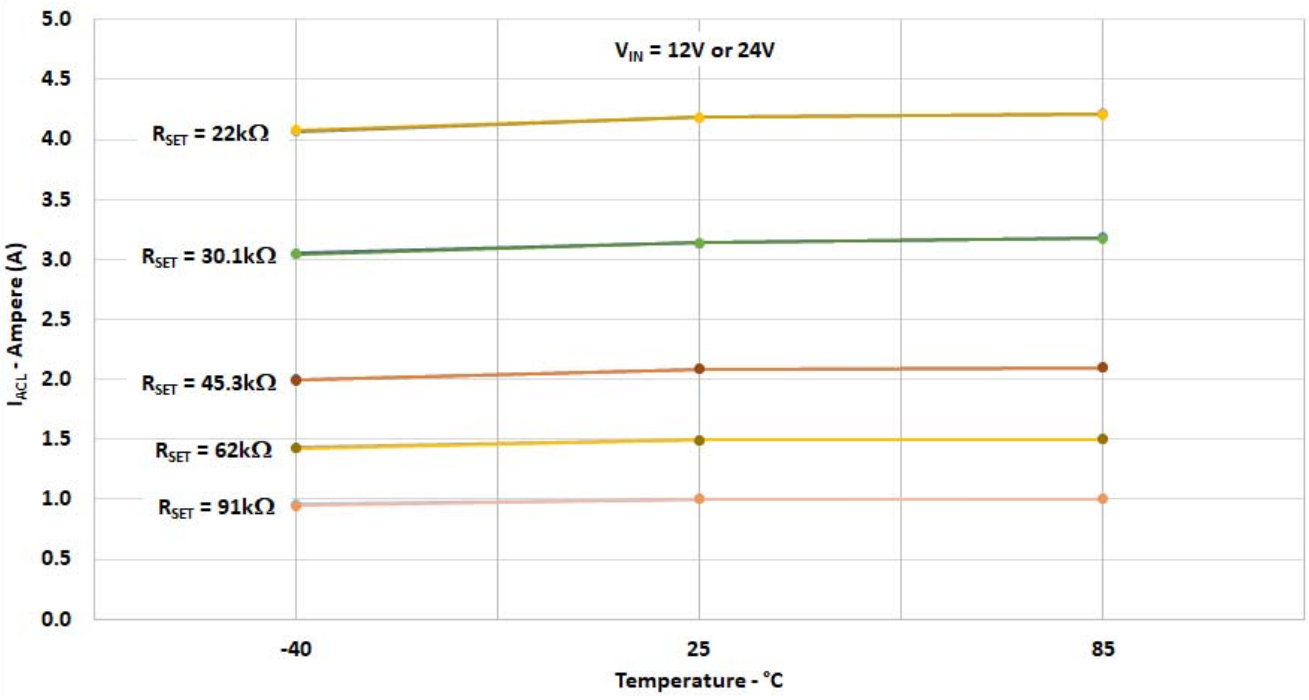
*Rise and Fall Times of the ON Signal are 100 ns

Typical Performance Characteristics

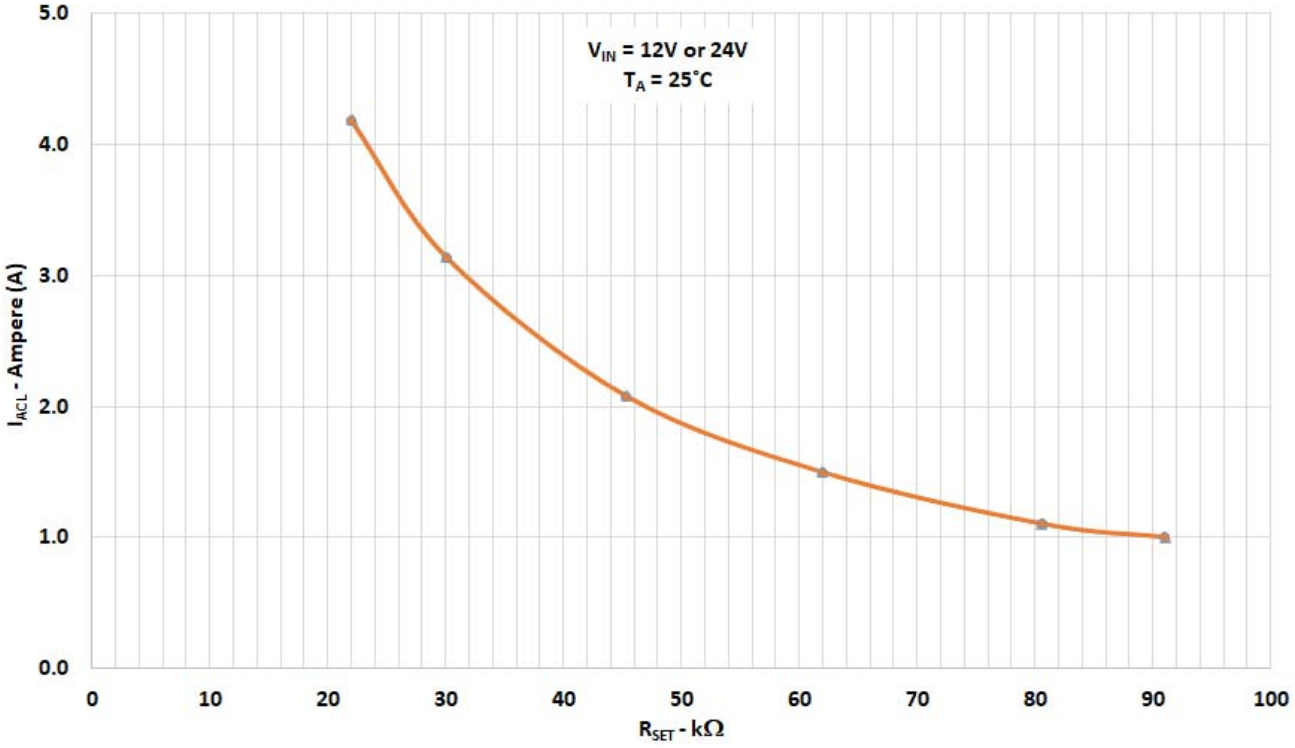
RDS_{ON} vs. Temperature and V_{IN}



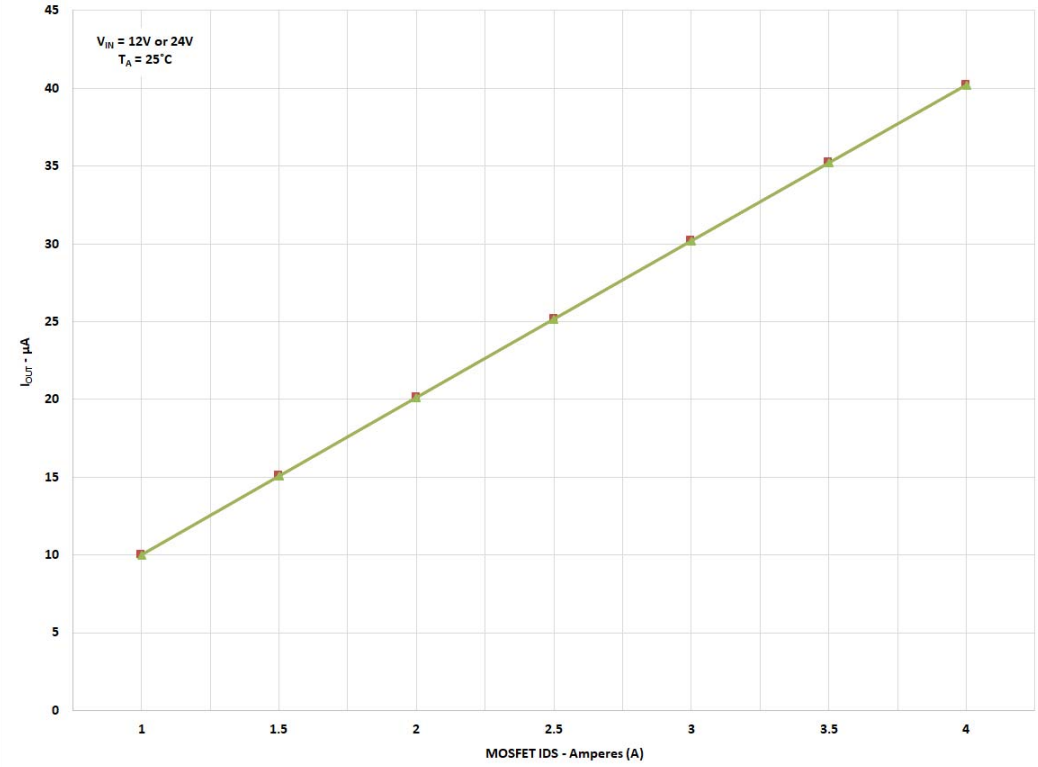
I_{ACL} vs. Temperature and R_{SET}



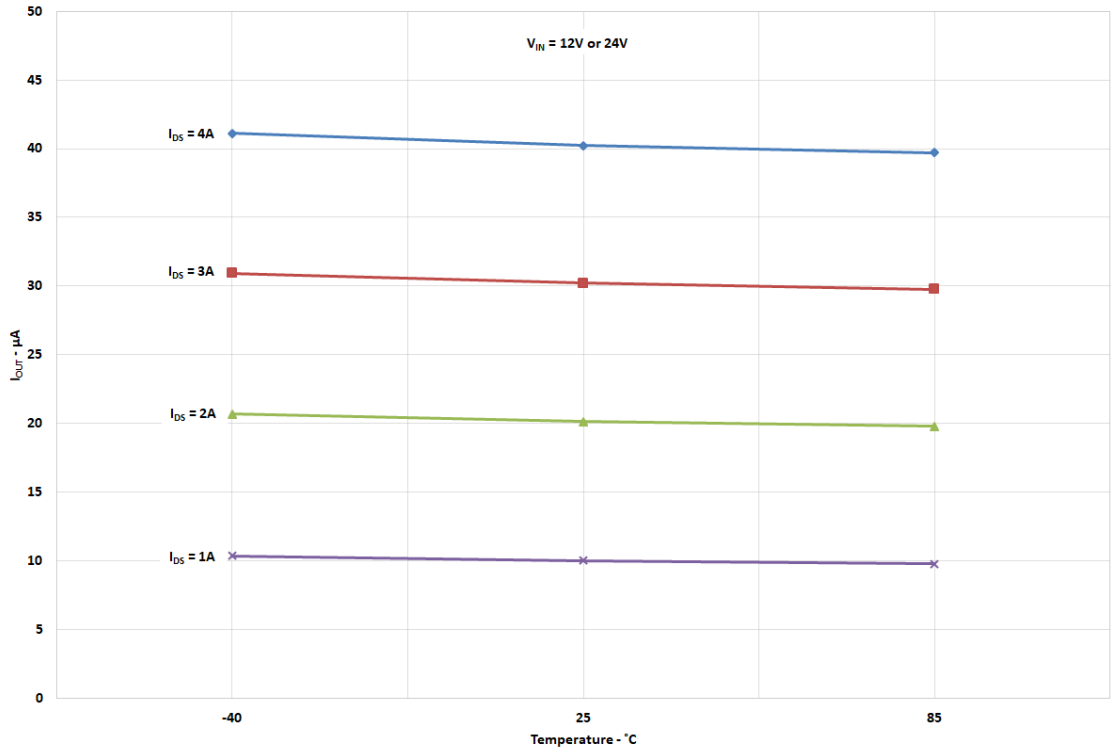
I_{ACL} vs. R_{SET} and V_{IN}



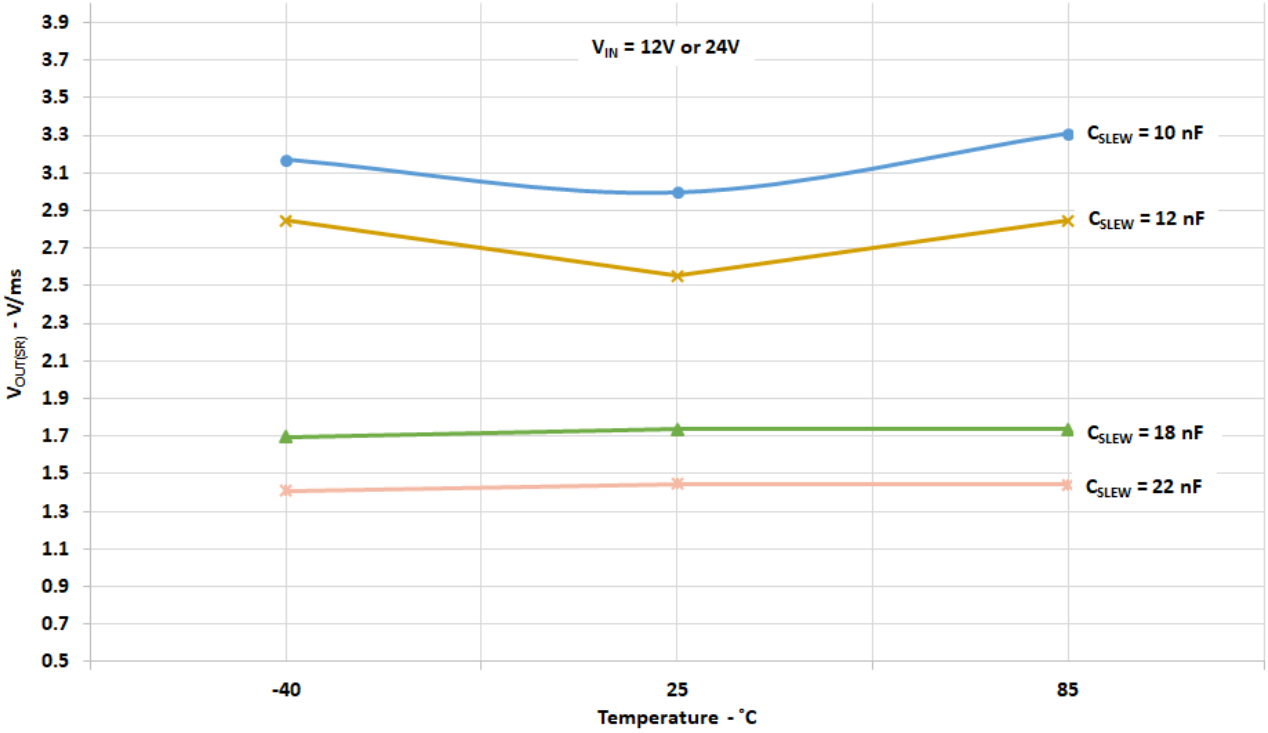
I_{OUT} vs. MOSFET IDS and V_{IN}



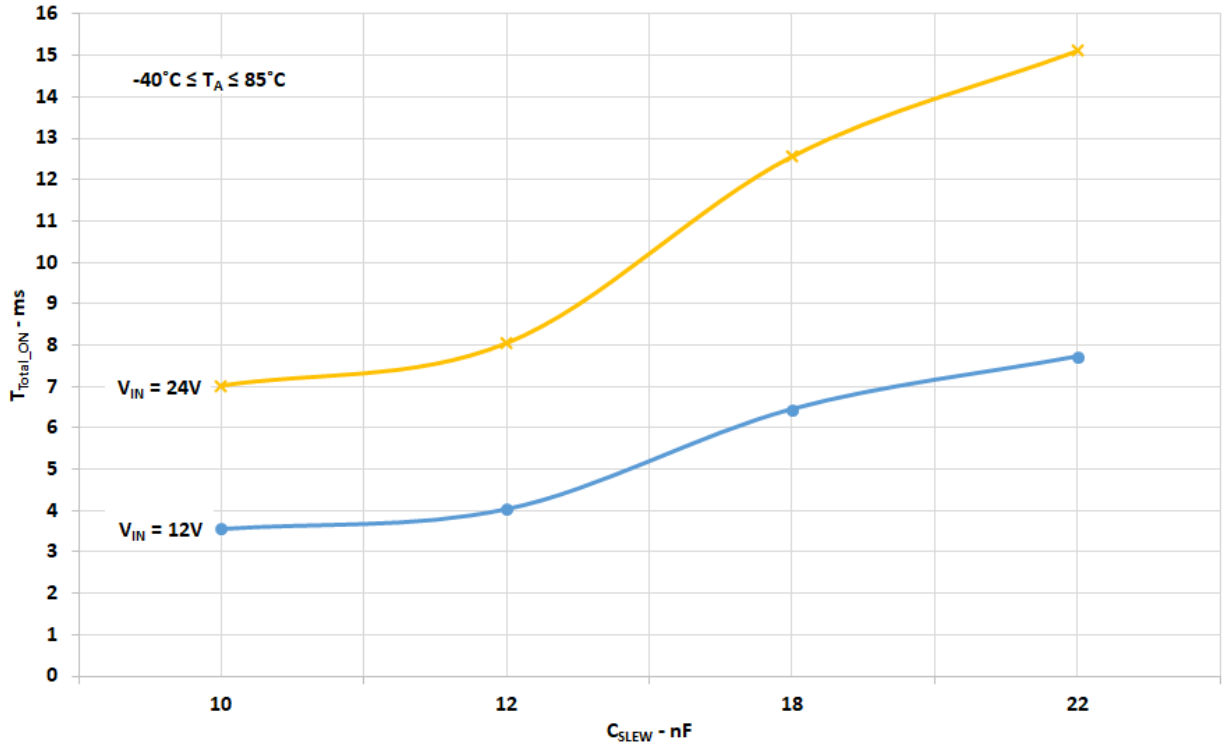
I_{OUT} vs. Temperature and MOSFET IDS



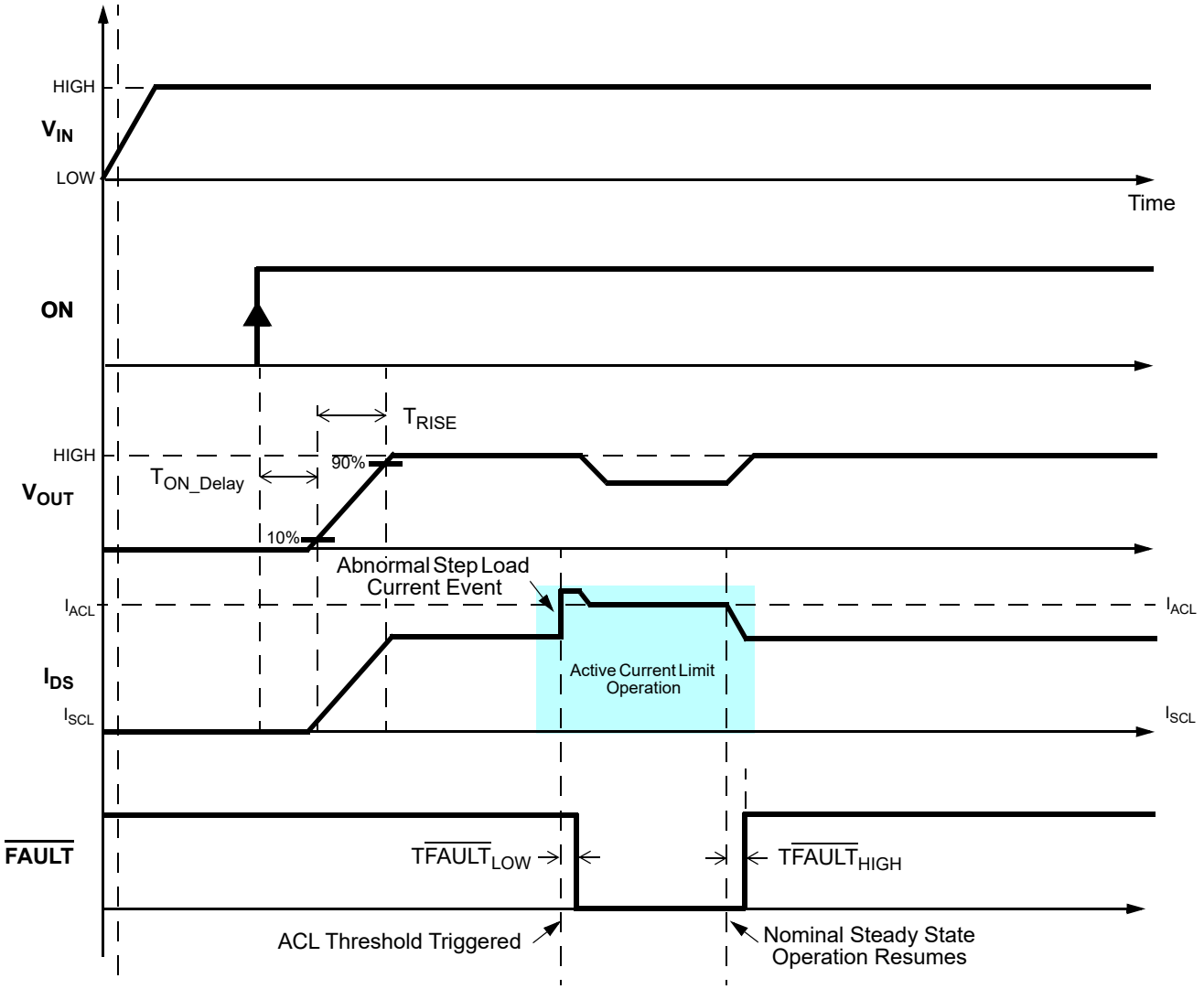
V_{OUT} Slew Rate vs. Temperature, V_{IN}, and C_{SLEW}



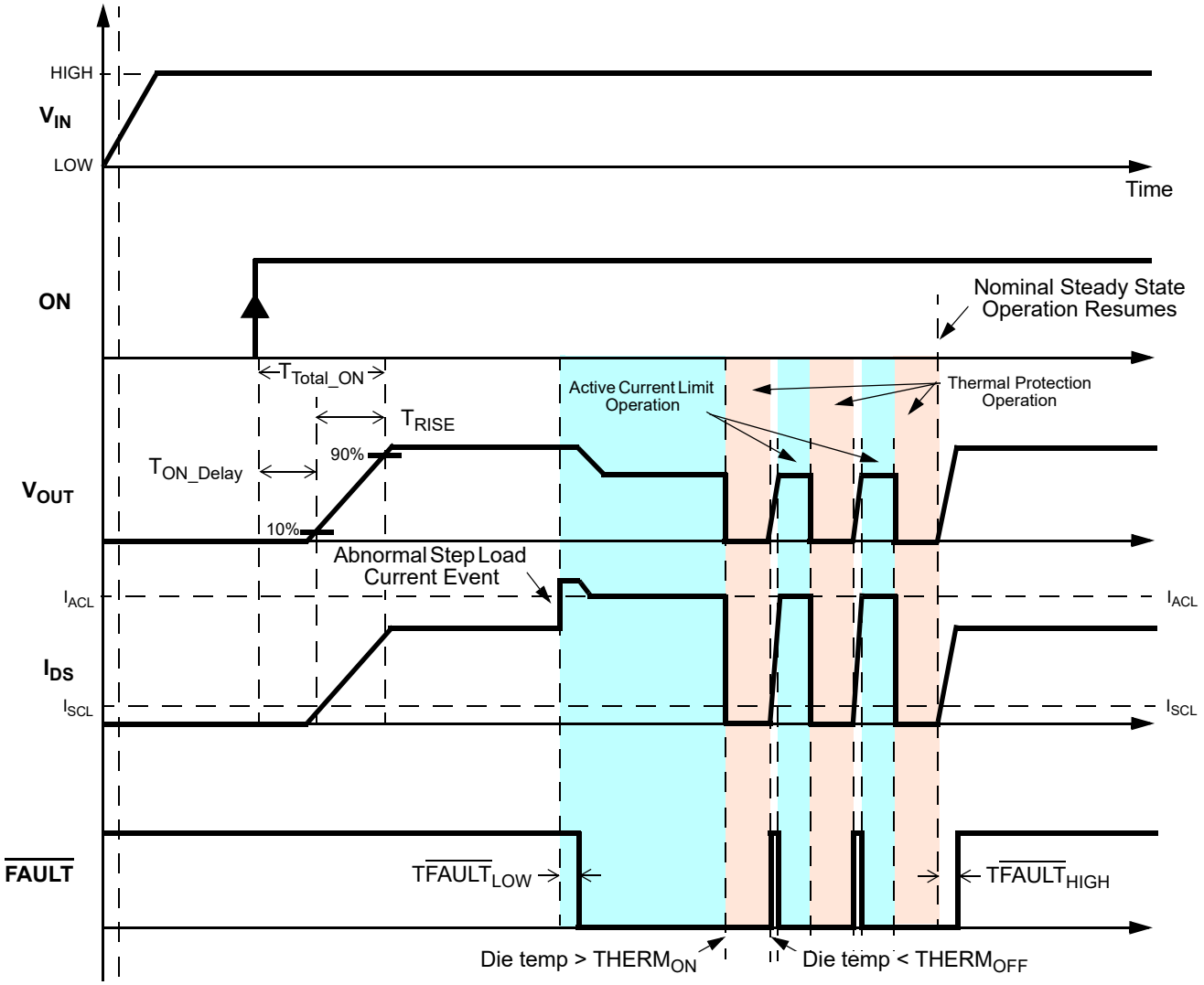
T_{Total_ON} vs. C_{SLEW} and V_{IN}



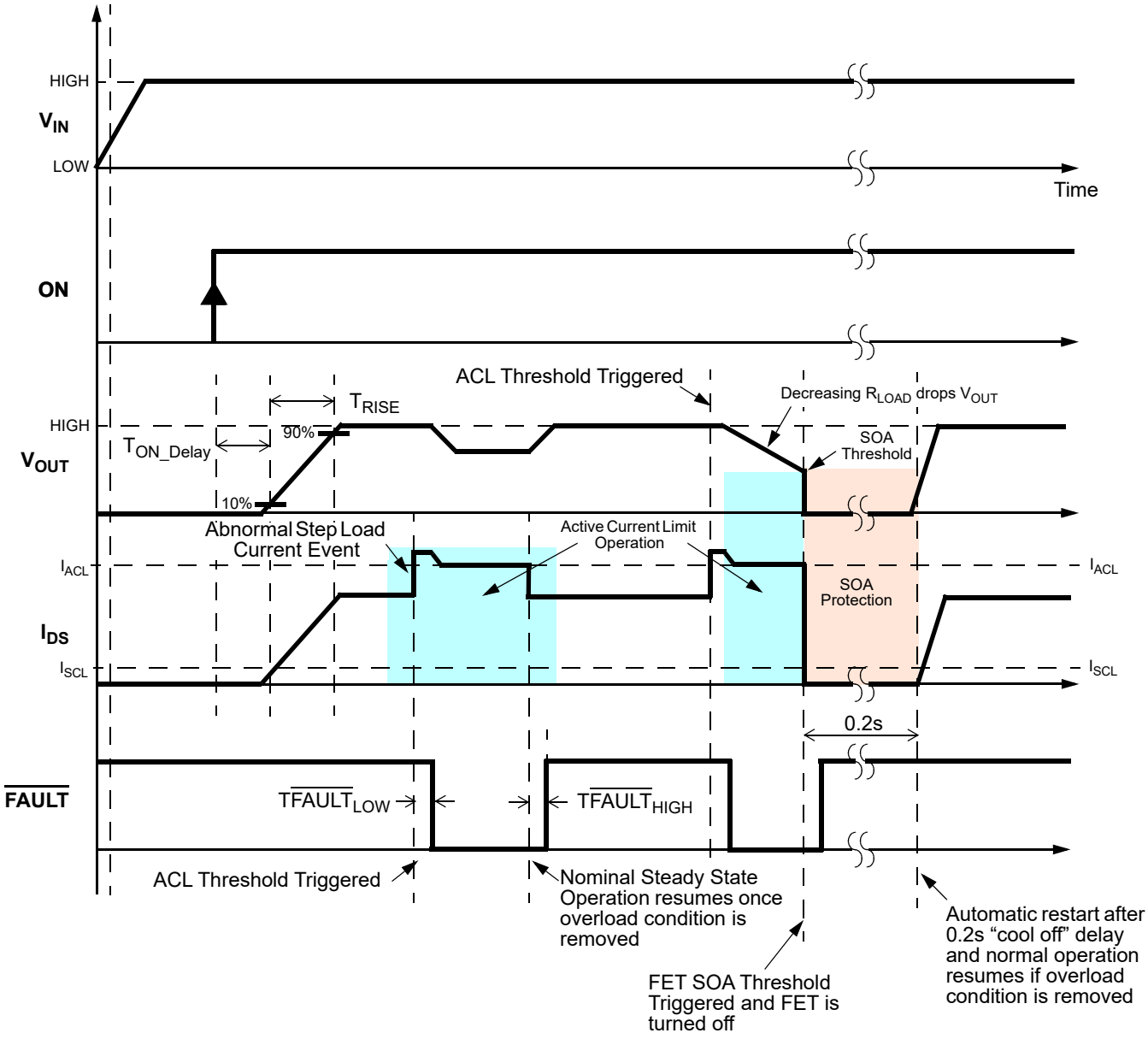
Timing Diagram - Basic Operation including Active Current Limit Protection



Timing Diagram - Active Current Limit & Thermal Protection Operation



Timing Diagram - Basic Operation including Active Current + Internal FET SOA Protection



SLG59H1017V Application Diagram

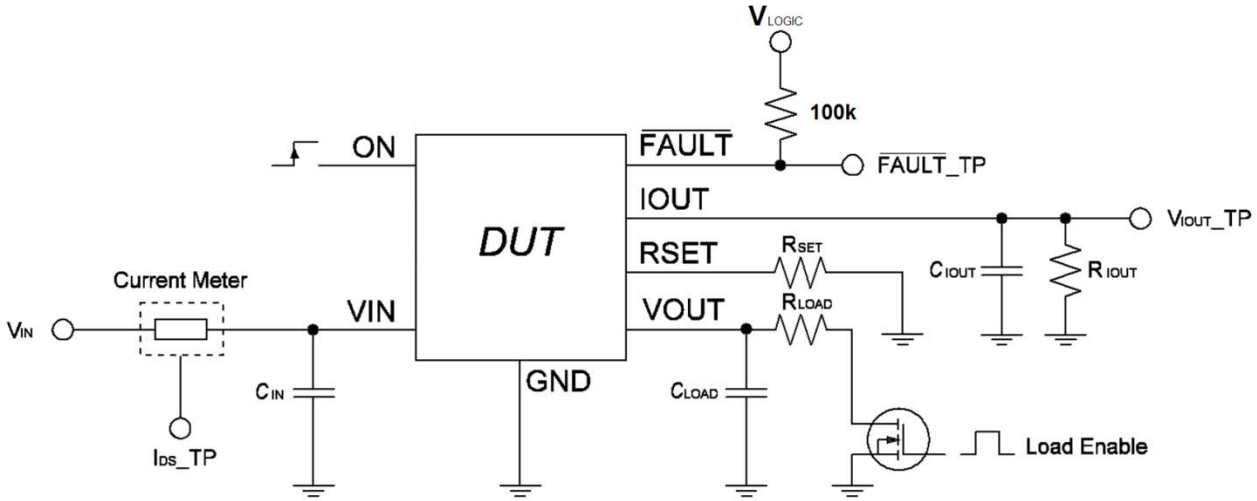


Figure 1. Test setup Application Diagram

Typical Turn-on Waveforms

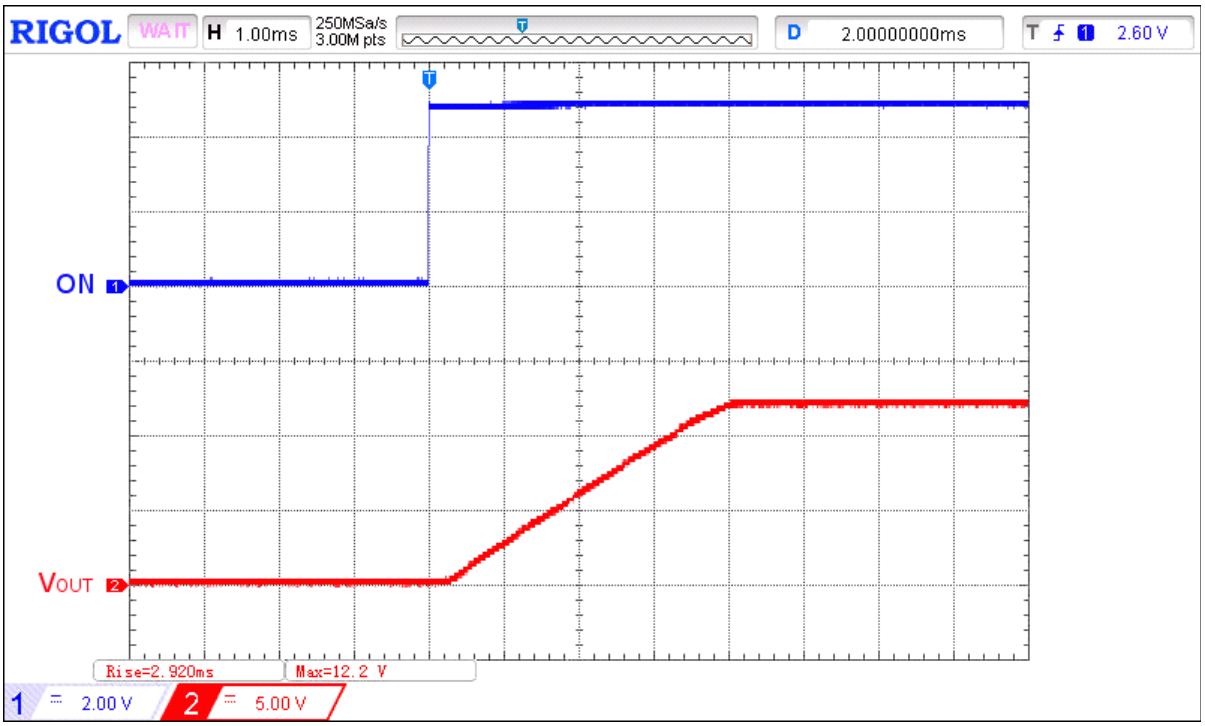


Figure 2. Typical Turn ON operation waveform for V_{IN} = 12 V, C_{SLEW} = 10 nF, C_{LOAD} = 10 µF, R_{LOAD} = 100 Ω

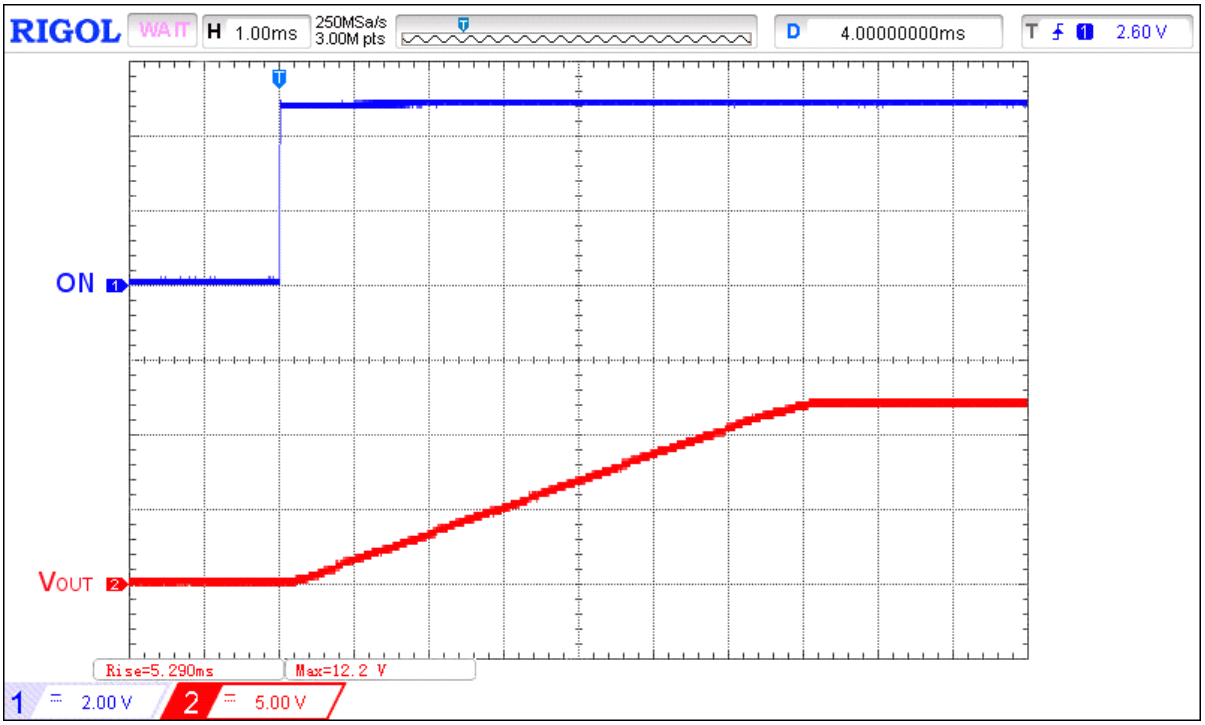


Figure 3. Typical Turn ON operation waveform for V_{IN} = 12 V, C_{SLEW} = 18 nF, C_{LOAD} = 10 µF, R_{LOAD} = 100 Ω

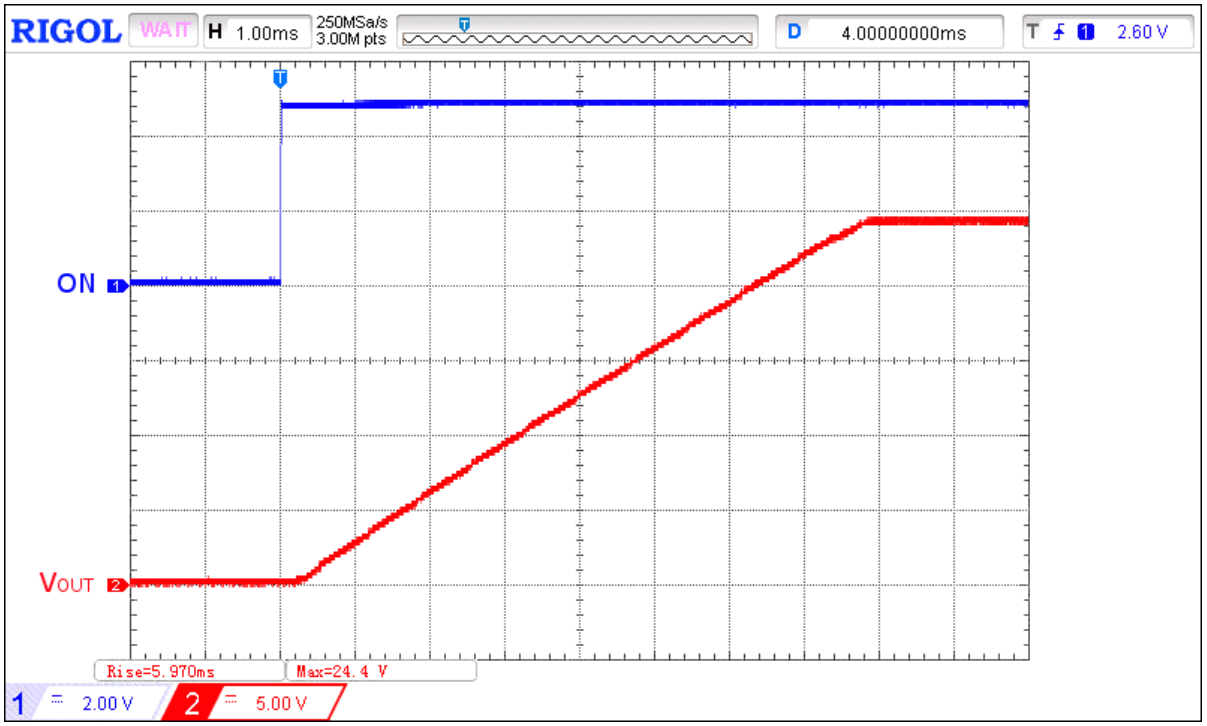


Figure 4. Typical Turn ON operation waveform for V_{IN} = 24 V, C_{SLEW} = 10 nF, C_{LOAD} = 10 µF, R_{LOAD} = 100 Ω

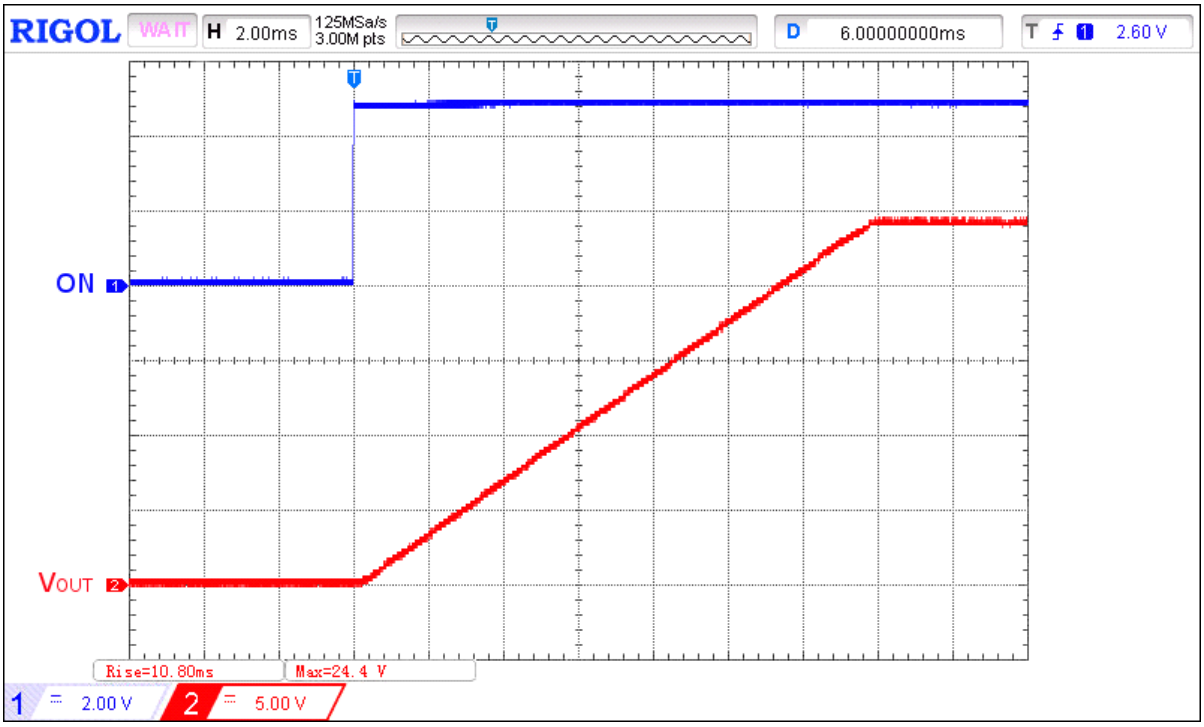


Figure 5. Typical Turn ON operation waveform for V_{IN} = 24 V, C_{SLEW} = 18 nF, C_{LOAD} = 10 µF, R_{LOAD} = 100 Ω

Typical Turn-off Waveforms

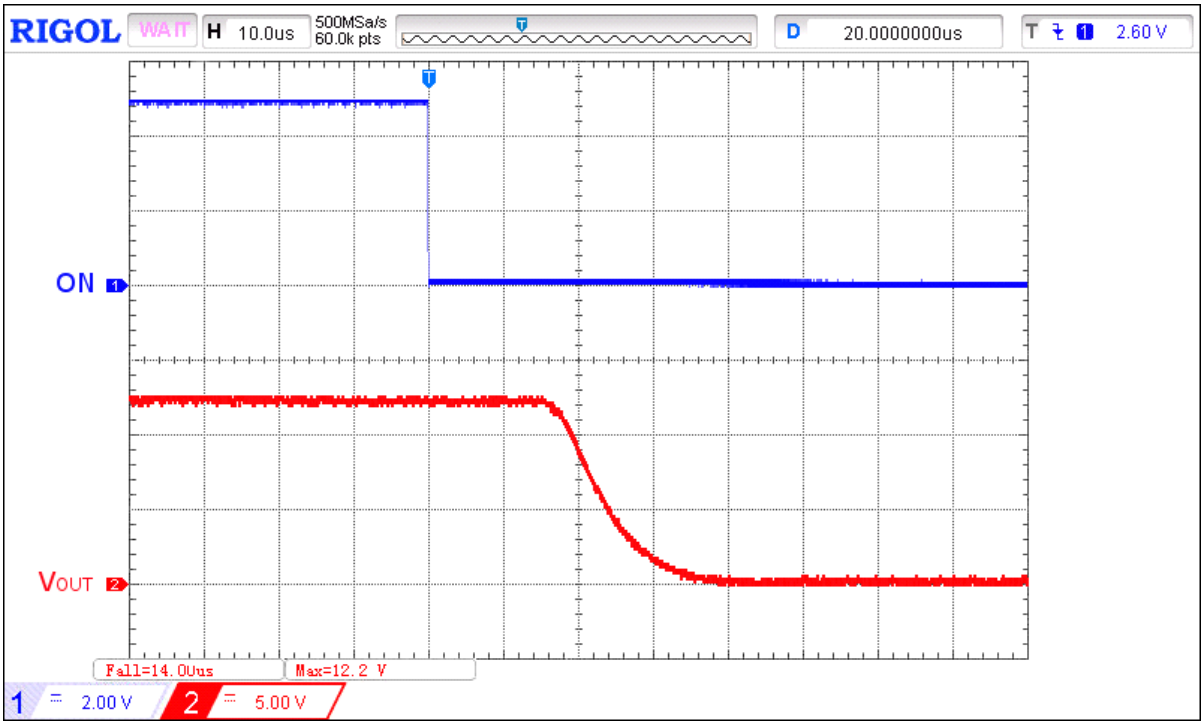


Figure 6. Typical Turn OFF operation waveform for V_{IN} = 12 V, C_{SLEW} = 10 nF, no C_{LOAD} , R_{LOAD} = 100 Ω

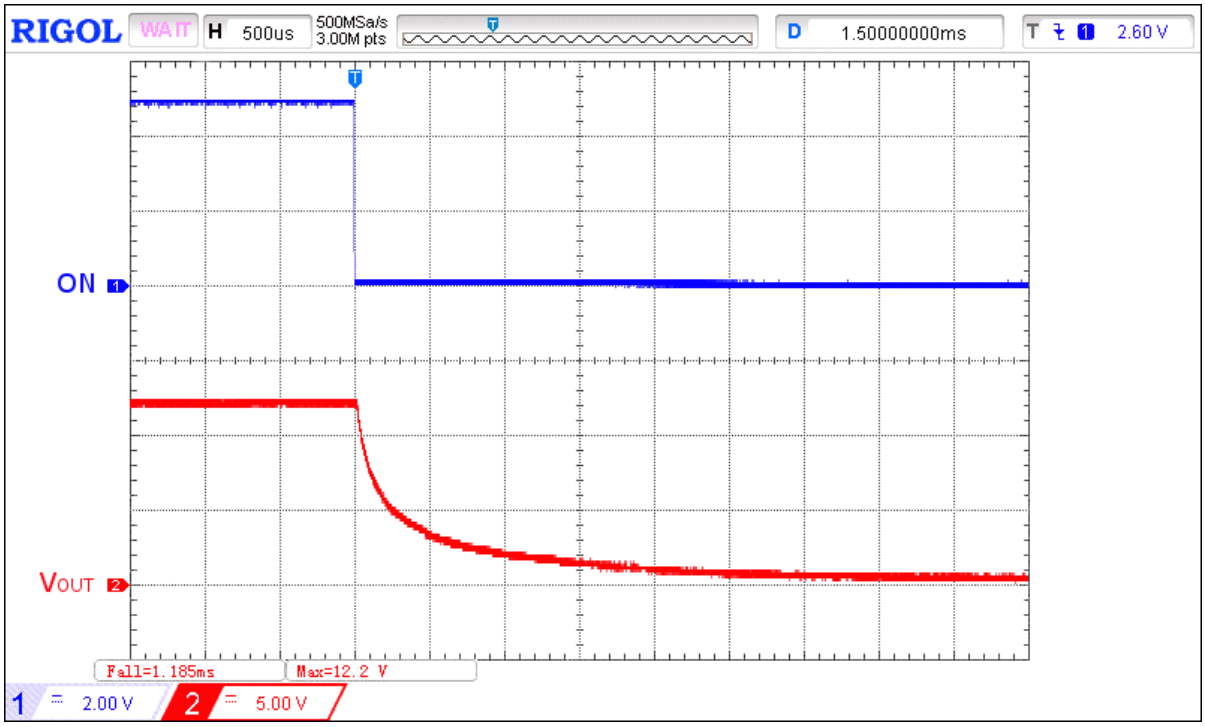


Figure 7. Typical Turn OFF operation waveform for V_{IN} = 12 V, C_{SLEW} = 10 nF, C_{LOAD} = 10 µF, R_{LOAD} = 100 Ω

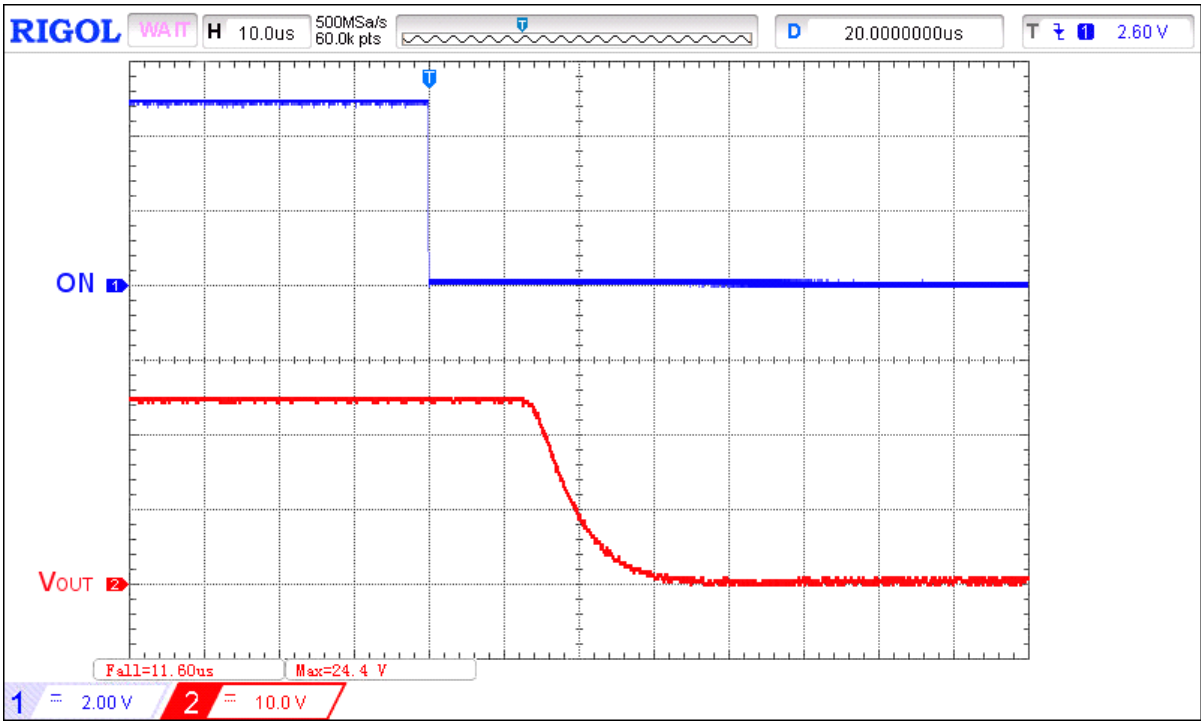


Figure 8. Typical Turn OFF operation waveform for V_{IN} = 24 V, C_{SLEW} = 10 nF, no C_{LOAD} , R_{LOAD} = 100 Ω

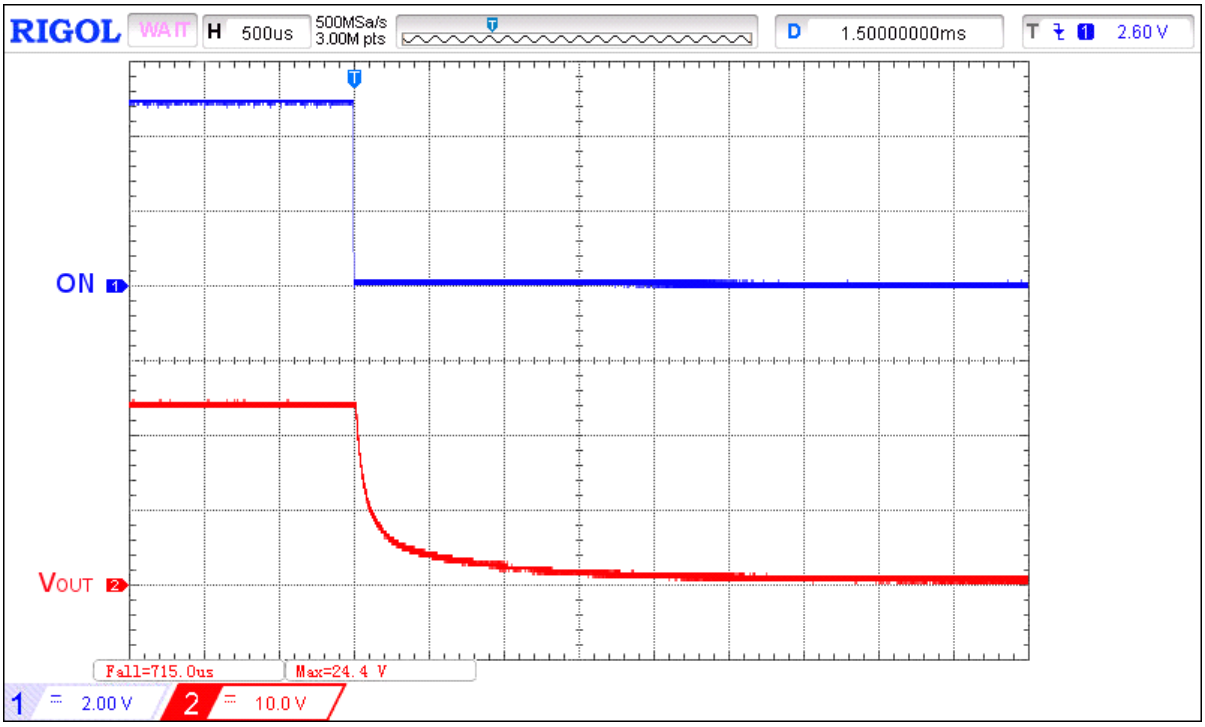


Figure 9. Typical Turn OFF operation waveform for V_{IN} = 24 V, C_{SLEW} = 10 nF, C_{LOAD} = 10 µF, R_{LOAD} = 100 Ω

Typical ACL Operation Waveforms

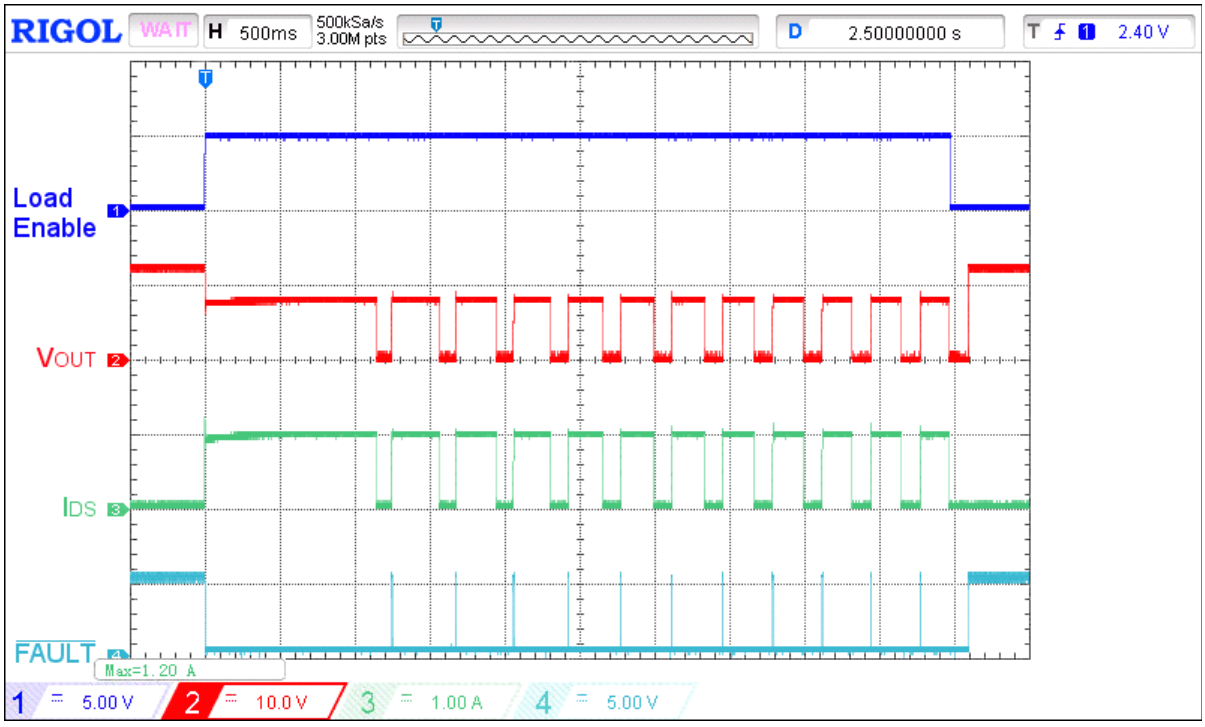


Figure 10. Typical ACL operation waveform for V_{IN} = 12 V, C_{LOAD} = 10 µF, I_{ACL} = 1 A, R_{SET} = 91 kΩ

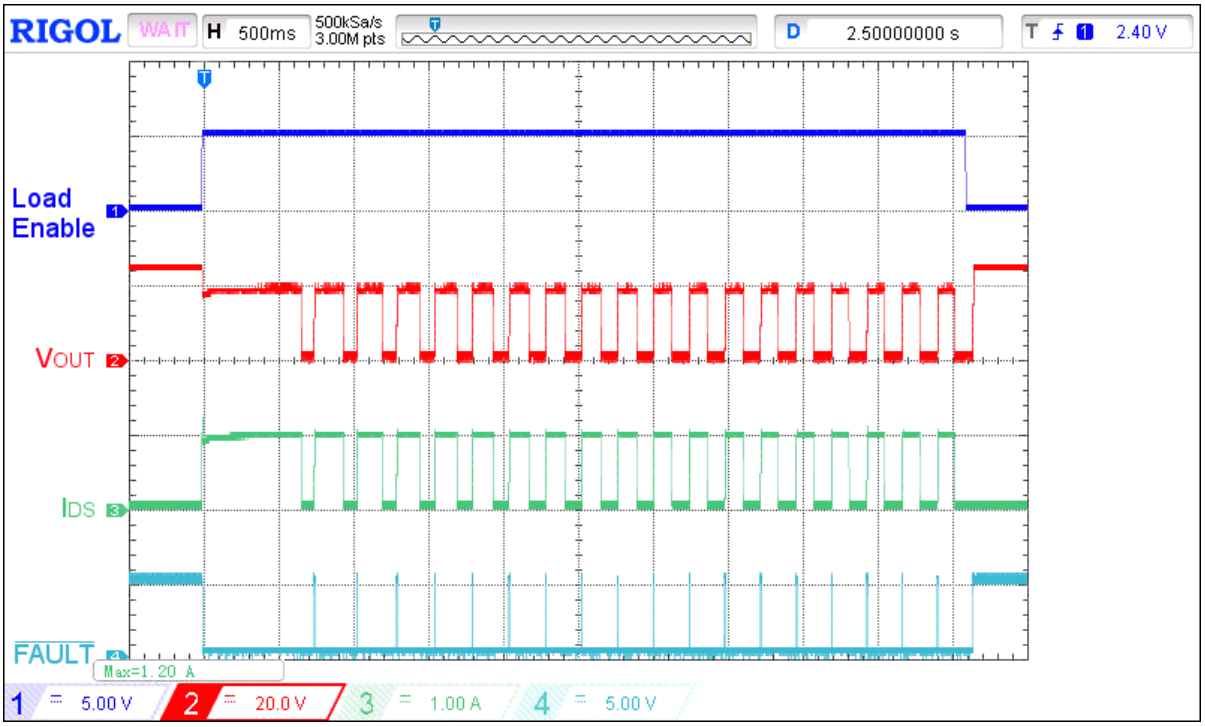


Figure 11. Typical ACL operation waveform for V_{IN} = 24 V, C_{LOAD} = 10 µF, I_{ACL} = 1 A, R_{SET} = 91 kΩ

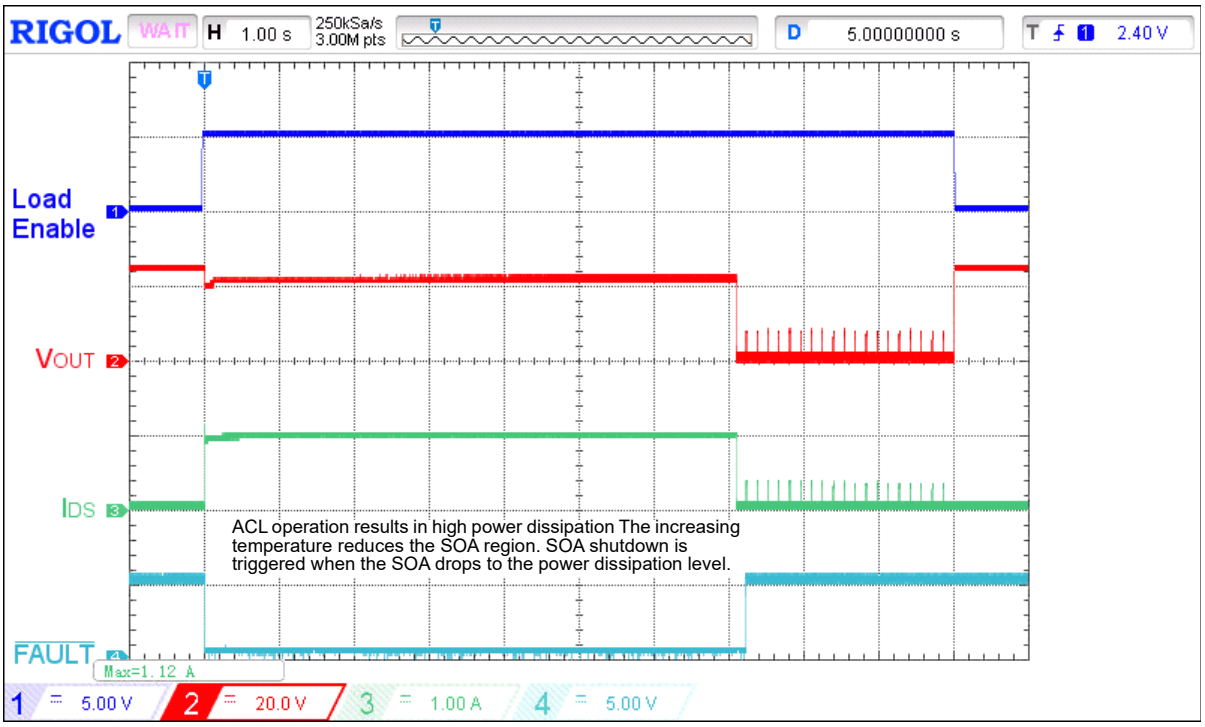


Figure 12. Thermally induced SOA shutdown for V_{IN} = 24 V, C_{LOAD} = 10 µF, R_{SET} = 91 kΩ, I_{ACL} = 1 A, R_{LOAD} = 20 Ω,

Typical $\overline{\text{FAULT}}$ Operation Waveforms

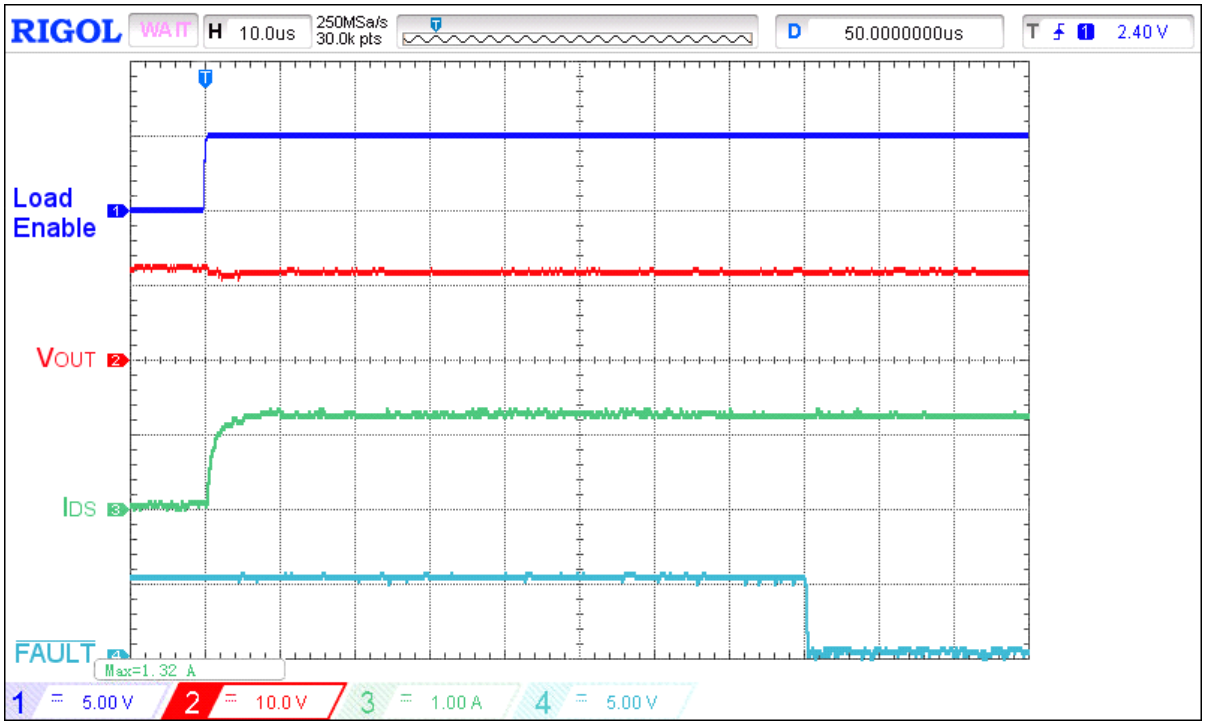


Figure 13. Typical $\overline{\text{FAULT}}$ assertion waveform for V_{IN} = 12 V, C_{LOAD} = 10 µF, I_{ACL} = 1 A, R_{SET} = 91 kΩ, switch on 9 Ω load

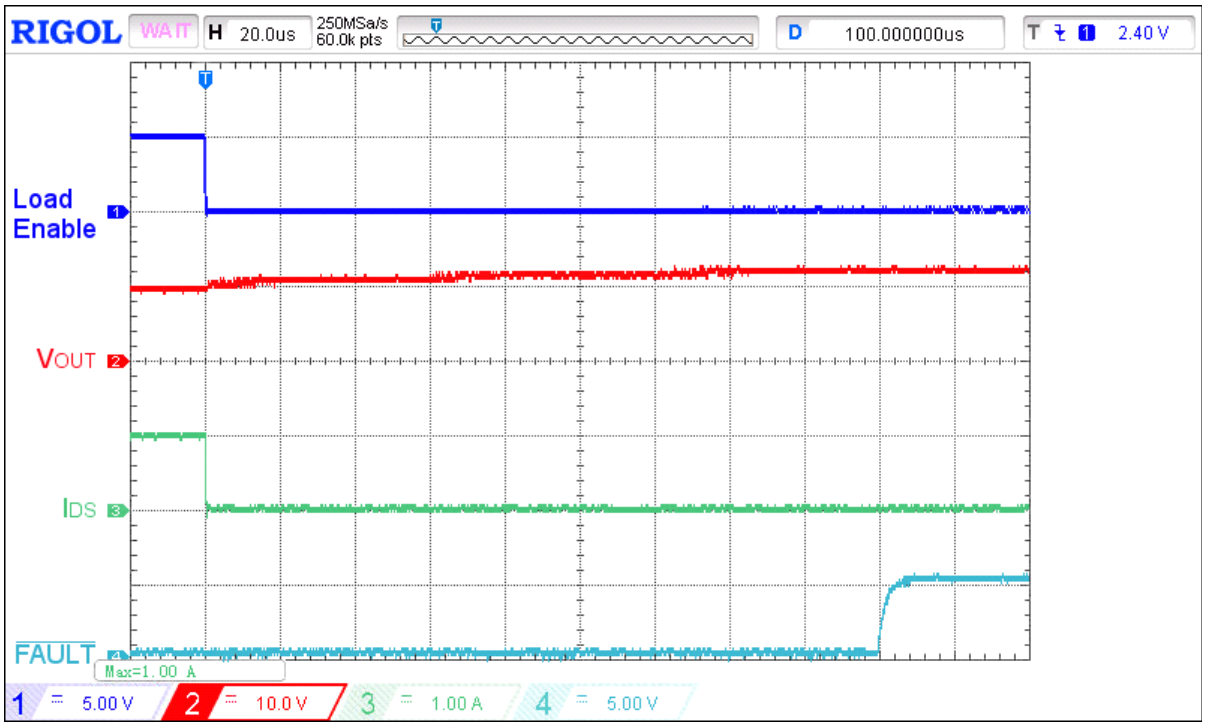


Figure 14. Typical $\overline{\text{FAULT}}$ de-assertion waveform for V_{IN} = 12 V, C_{LOAD} = 10 µF, I_{ACL} = 1 A, R_{SET} = 91 kΩ, switch out 9 Ω load

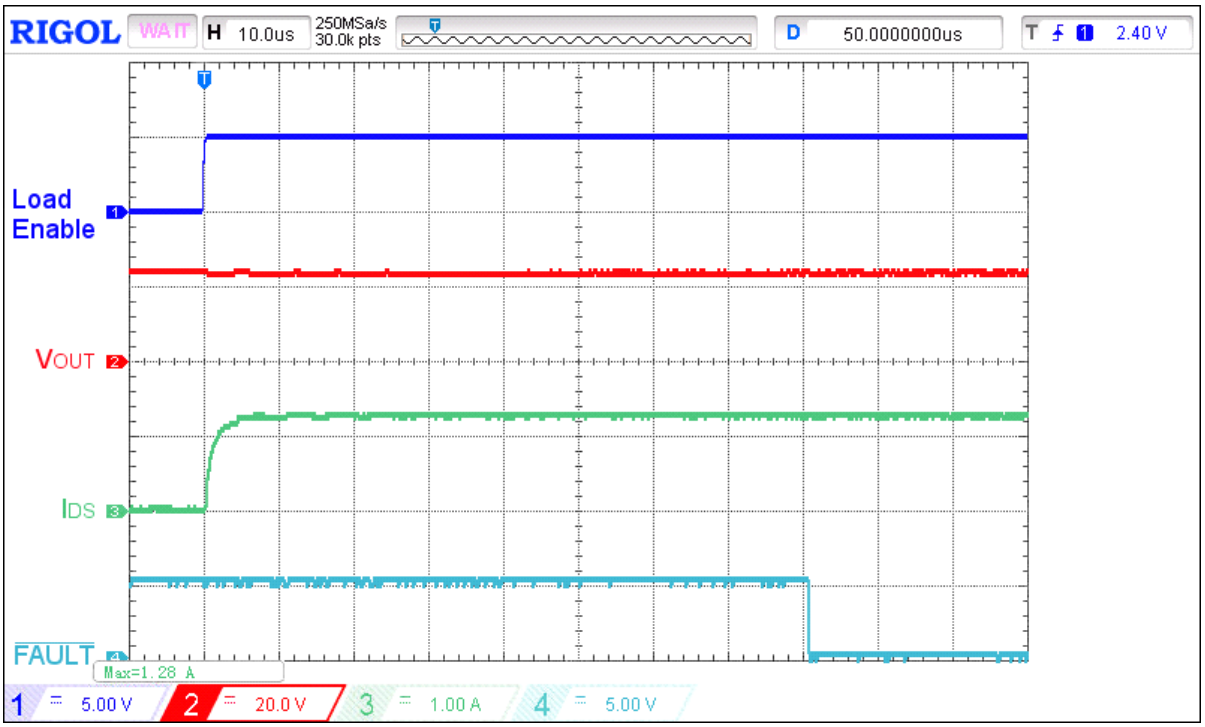


Figure 15. Typical $\overline{\text{FAULT}}$ assertion waveform for V_{IN} = 24 V, C_{LOAD} = 10 µF, I_{ACL} = 1 A, R_{SET} = 91 kΩ, switch on 18.5 Ω load

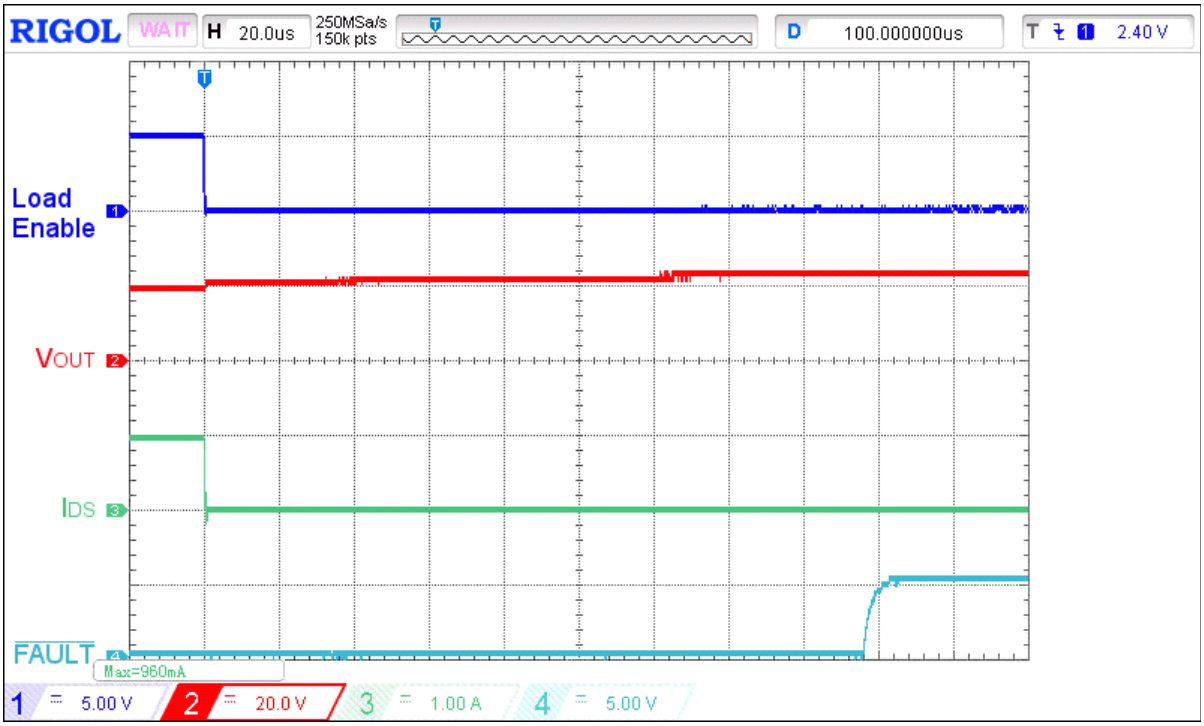


Figure 16. Typical $\overline{\text{FAULT}}$ de-assertion waveform for V_{IN} = 24 V, C_{LOAD} = 10 µF, I_{ACL} = 1 A, R_{SET} = 91 kΩ, switch out 18.5 Ω load

Typical I_{OUT} Response Time Waveforms

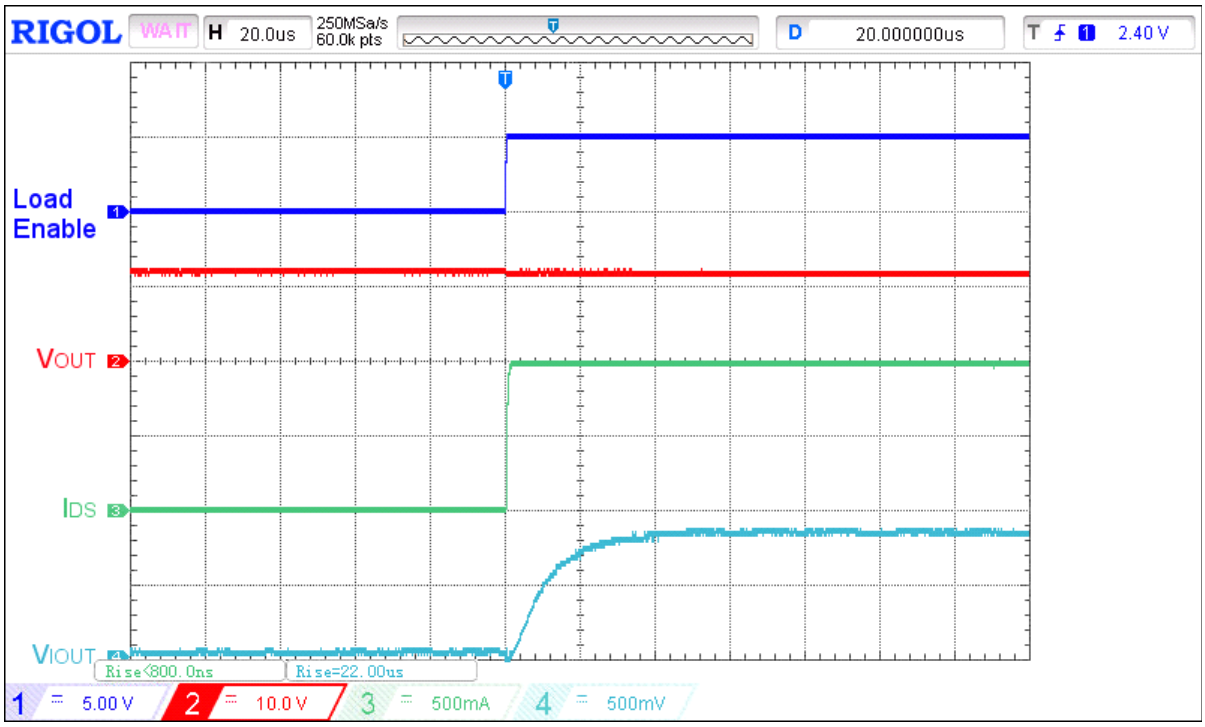


Figure 17. Typical I_{OUT} response time waveform for V_{IN} = 12 V, C_{LOAD} = 10 µF, R_{LOAD} = 12 Ω C_{IOUT} = 0.18 nF, R_{IOUT} = 84.5 kΩ, Load step 0 A to 1 A

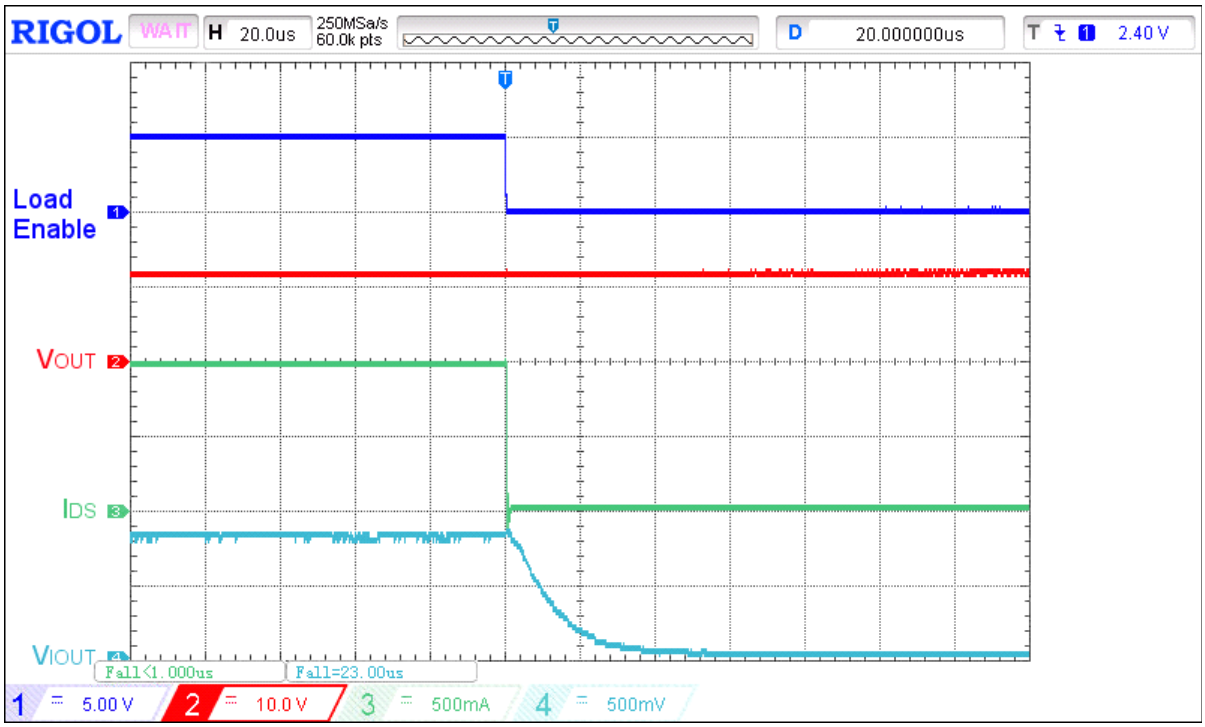


Figure 18. Typical I_{OUT} response time waveform for V_{IN} = 12 V, C_{LOAD} = 10 µF, R_{LOAD} = 12 Ω C_{IOUT} = 0.18 nF, R_{IOUT} = 84.5 kΩ, Load step 1 A to 0 A

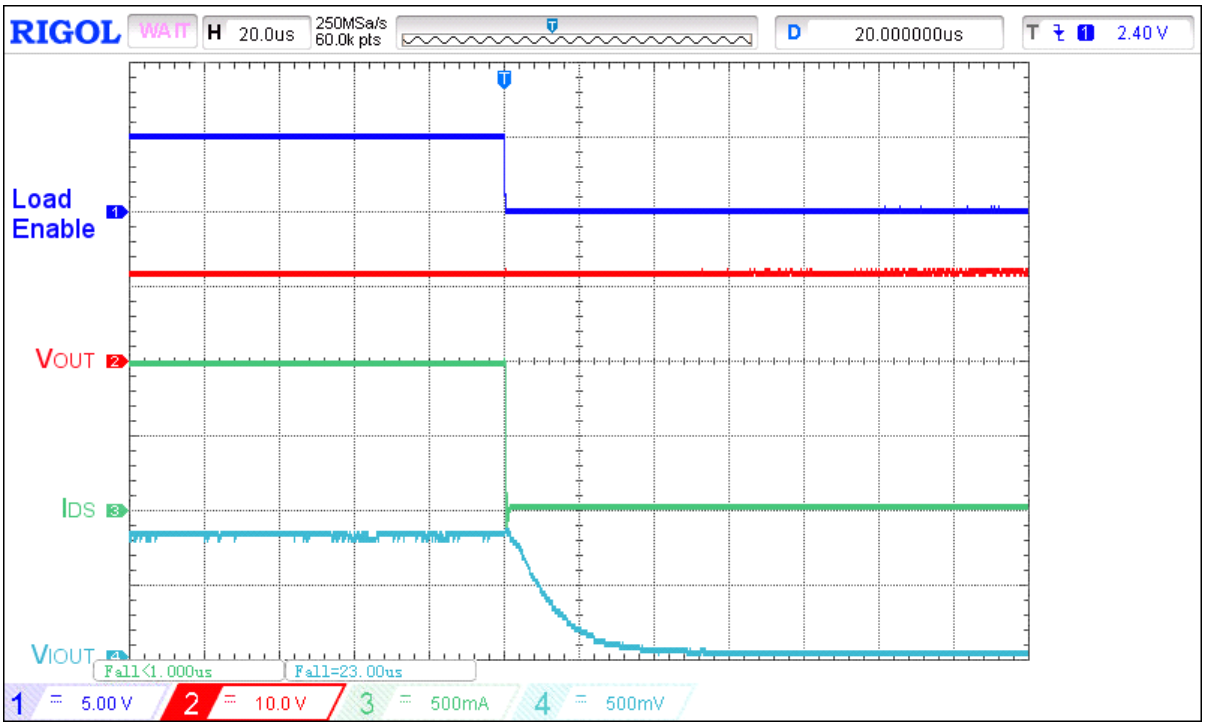


Figure 19. Typical I_{OUT} response time waveform for V_{IN} = 24 V, C_{LOAD} = 10 µF, R_{LOAD} = 24 Ω C_{IOUT} = 0.18 nF, R_{IOUT} = 84.5 kΩ, Load step 0 A to 1 A

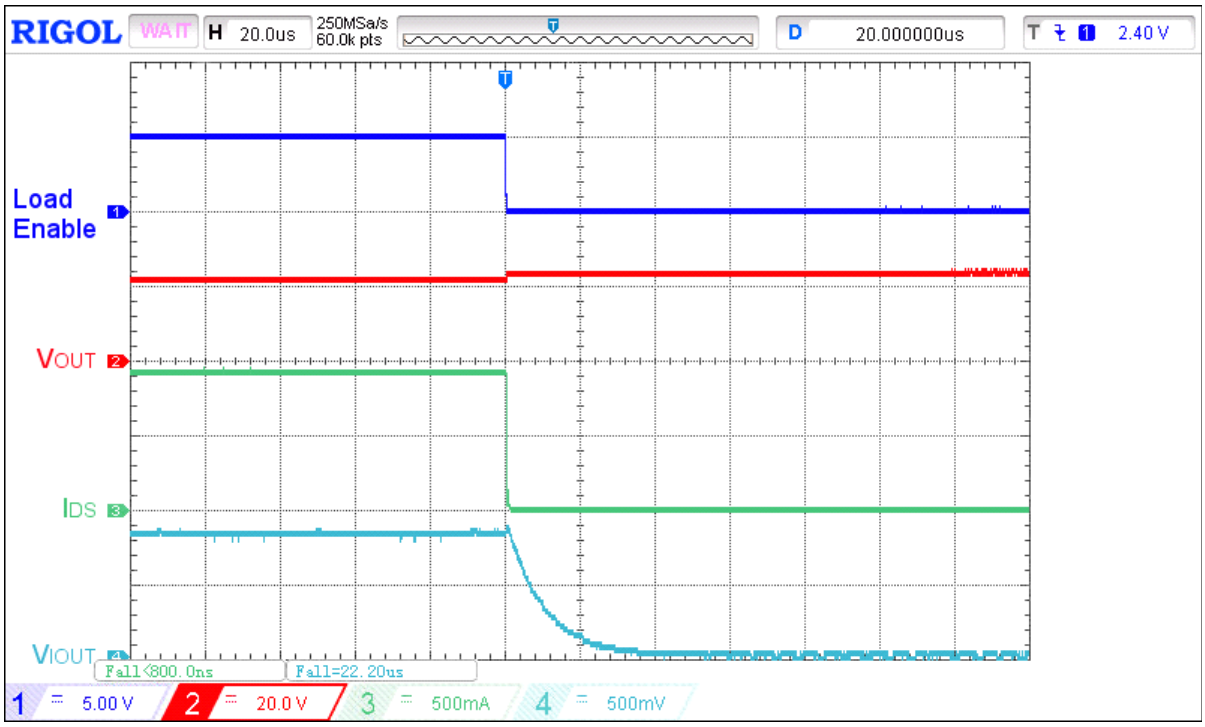


Figure 20. Typical I_{OUT} response time waveform for V_{IN} = 24 V, C_{LOAD} = 10 µF, R_{LOAD} = 24 Ω C_{IOUT} = 0.18 nF, R_{IOUT} = 84.5 kΩ, Load step 1 A to 0 A

Typical SOA Waveforms

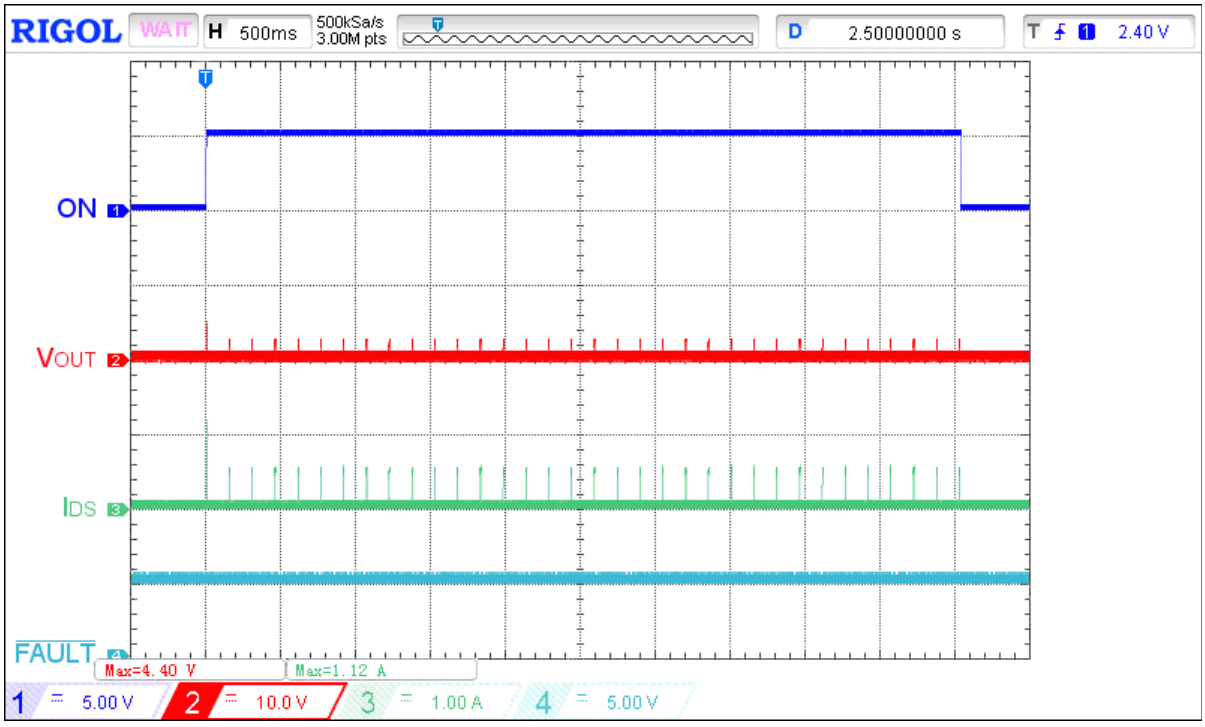


Figure 21. Typical SOA waveform during power up on heavy load for V_{IN} = 12 V, C_{LOAD} = 10 µF, R_{SET} = 30.1 kΩ, R_{LOAD} = 3.5 Ω

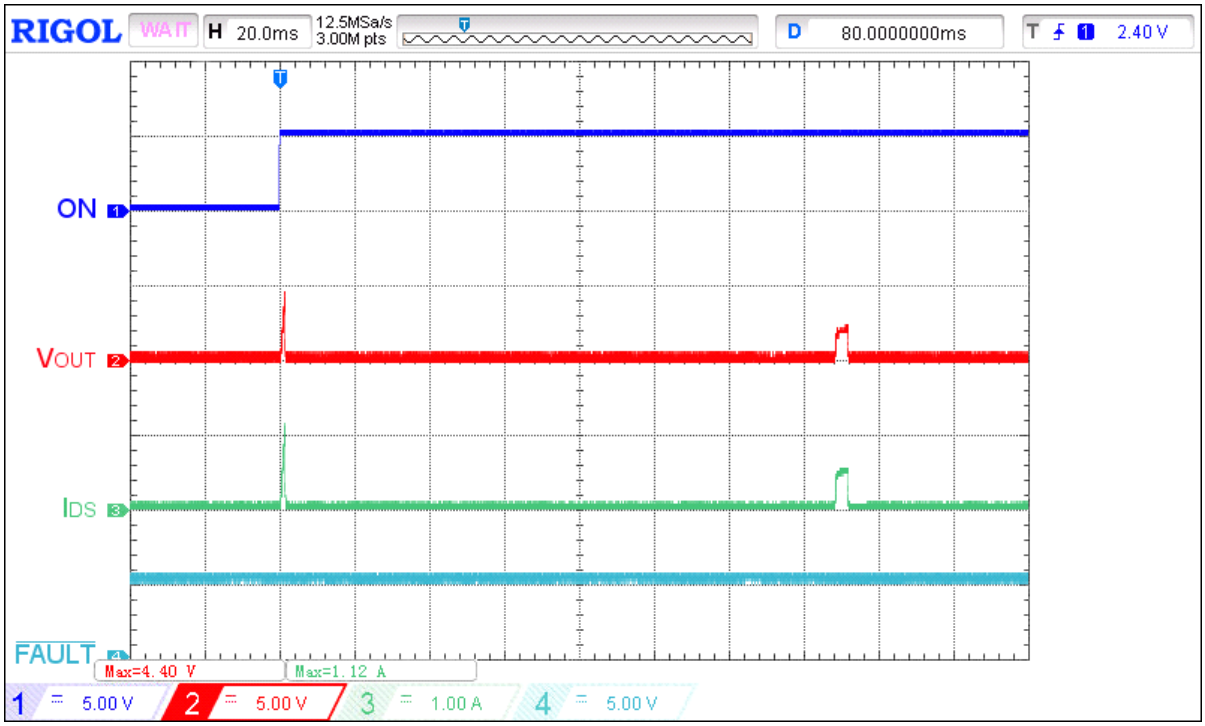


Figure 22. Extended typical SOA waveform during power up under heavy load for V_{IN} = 12 V, C_{LOAD} = 10 µF, R_{SET} = 30.1 kΩ, R_{LOAD} = 3.5 Ω

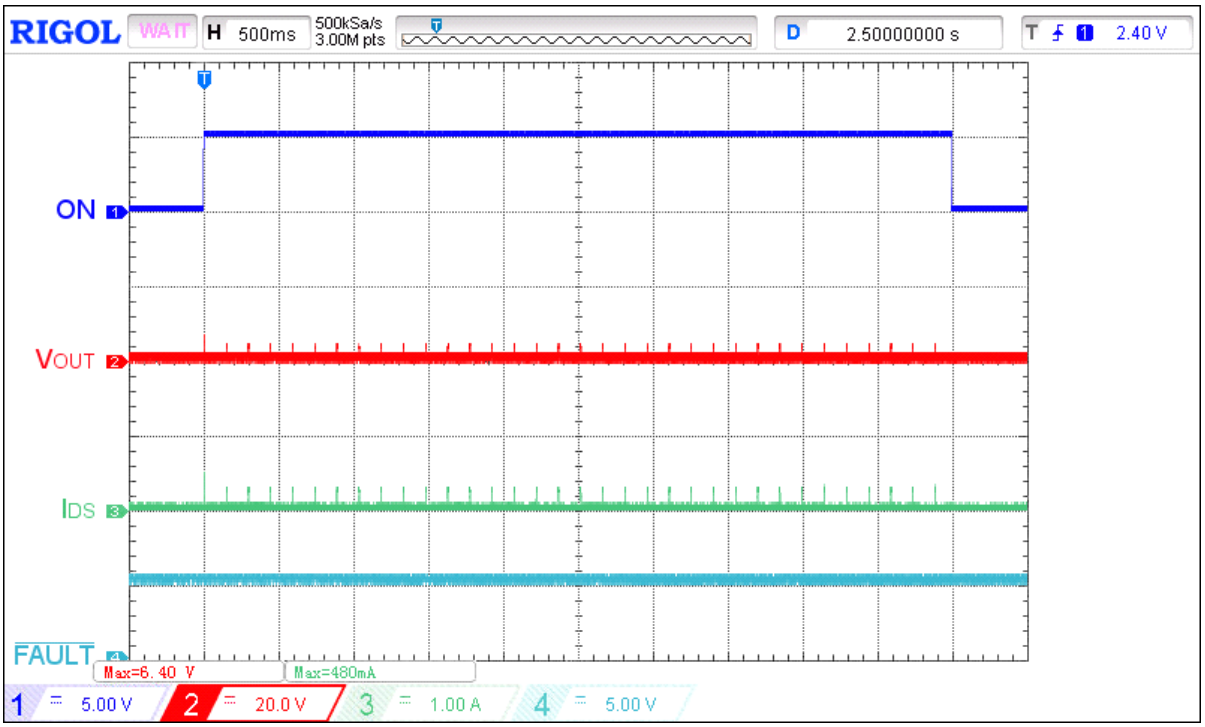


Figure 23. Typical SOA waveform during power up under heavy load for V_{IN} = 24 V, C_{LOAD} = 10 µF, R_{SET} = 30.1 kΩ, R_{LOAD} = 12 Ω

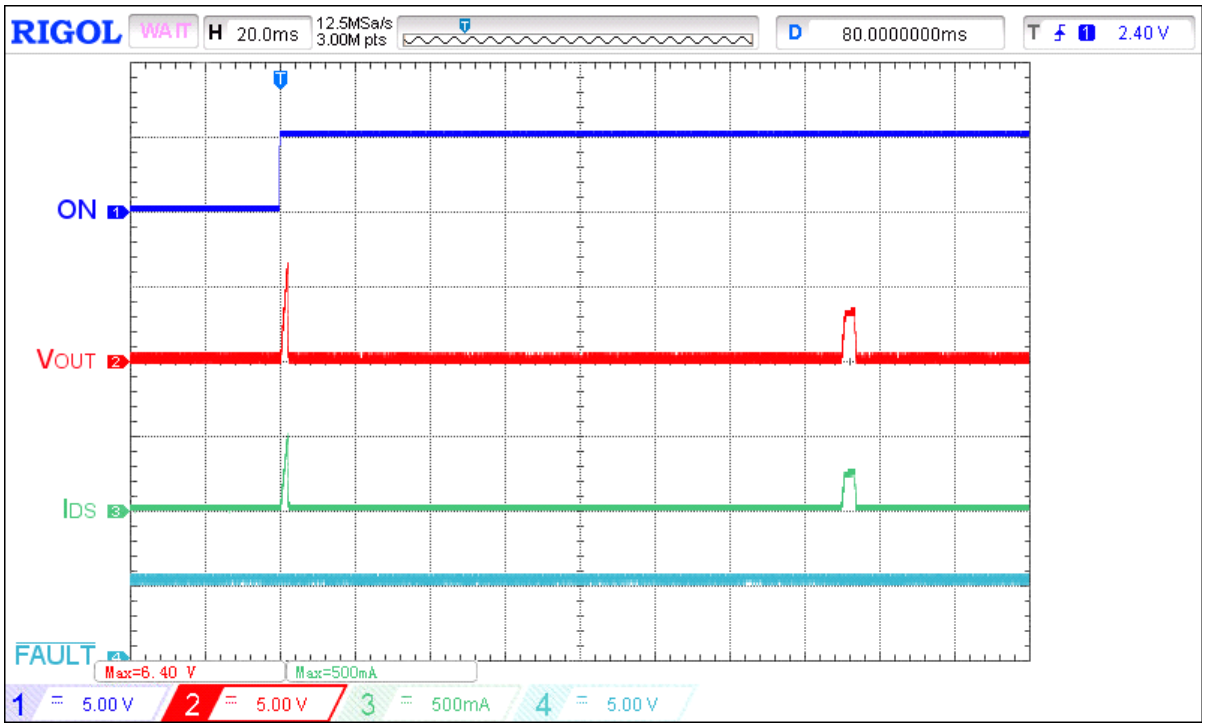


Figure 24. Extended typical SOA waveform during power up under heavy load for V_{IN} = 24 V, C_{LOAD} = 10 µF, R_{SET} = 30.1 kΩ, R_{LOAD} = 12 Ω

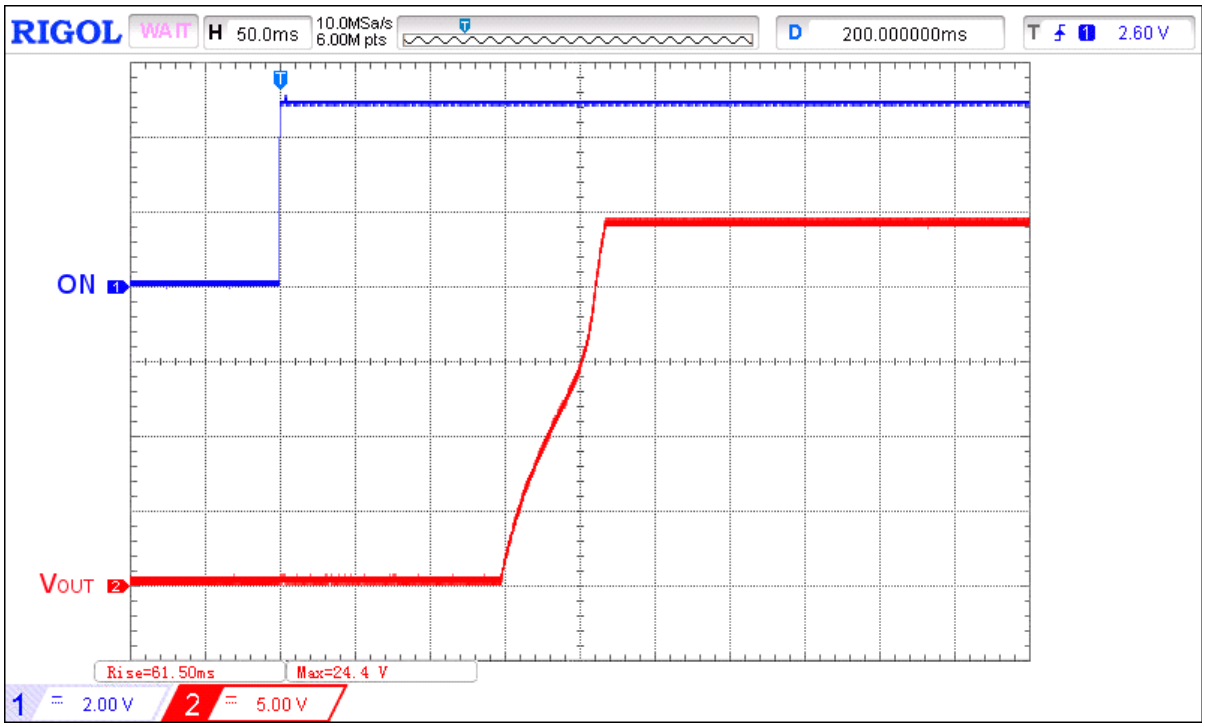


Figure 25. Typical non-monotonic V_{OUT} ramping waveform during power up on heavy load for V_{IN} = 24 V, C_{LOAD} = 470 µF, R_{SET} = 91 kΩ, R_{LOAD} = 40 Ω

Applications Information

HFET1 Safe Operating Area Explained

Dialog's HFET1 integrated power controllers incorporate a number of internal protection features that prevents them from damaging themselves or any other circuit or subcircuit downstream of them. One particular protection feature is their Safe Operation Area (SOA) protection. SOA protection is automatically activated under overpower and, in some cases, under overcurrent conditions. Overpower SOA is activated if package power dissipation exceeds an internal 10 W threshold and HFET1 devices will quickly switch off (open circuit) upon overpower detection and automatically resume (close) nominal operation once overpower condition no longer exists.

One of the possible ways to have an overpower condition trigger SOA protection is when HFET1 products are enabled into heavy output resistive loads and/or into large load capacitors. It is under these conditions to follow carefully the "Safe Start-up Loading" guidance in the Applications section of the datasheet. During an overcurrent condition, HFET1 devices will try to limit the output current to the level set by the external R_{SET} resistor. Limiting the output current, however, causes an increased voltage drop across the FET's channel because the FET's RDS_{ON} increased as well. Since the FET's RDS_{ON} is larger, package power dissipation also increases. If the resultant increase in package power dissipation is greater than or equal to 10 W, internal SOA protection will be triggered and the FET will open circuit (switch off). Every time SOA protection is triggered, all HFET1 devices will automatically attempt to resume nominal operation after 160 ms. The automatic retry attempt only allows power-up with SOA at 5 W. This SOA fold back power ensures that the FET survives a short circuit condition. To clear the 5 W SOA fold back, switch the ON pin to "LOW" to power reset SOA to 10 W.

Safe Start-up Condition

SLG59H1017V has built-in protection to prevent over-heating during start-up into a heavy load. Overloading the VOUT pin with a capacitor and a resistor may result in non-monotonic V_{OUT} ramping (*Figure 25*) or repeated restarts (*Figure 21* to *Figure 24*). In general, under light loading on VOUT, V_{OUT} ramping can be controlled with C_{SLEW} value. The following equation serves as a guide:

$$C_{SLEW} = \frac{T_{RISE}}{V_{IN}} \times 4.9\ \mu A \times \frac{20}{3}$$

where
T_{RISE} = Total rise time from 10% V_{OUT} to 90% V_{OUT}
V_{IN} = Input Voltage
C_{SLEW} = Capacitor value for CAP pin

When capacitor and resistor loading on VOUT during start up, the following tables will ensure V_{OUT} ramping is monotonic without triggering internal protection:

Safe Start-up Loading for V_{IN} = 12 V (Monotonic Ramp)			
Slew Rate (V/ms)	**C_{SLEW} (nF)[2]**	**C_{LOAD} (µF)**	**R_{LOAD} (Ω)**
1	33.3	500	8
2	16.7	250	8
3	11.1	160	8
4	8.3	120	8
5	6.7	100	8

Safe Start-up Loading for V_{IN} = 24 V (Monotonic Ramp)			
Slew Rate (V/ms)	C_{SLEW} (nF)[2]	C_{LOAD} (µF)	R_{LOAD} (Ω)
0.5	66.7	500	30
1.0	33.3	250	30
1.5	22.2	160	30
2.0	16.7	120	30
2.5	13.3	100	30

Note 2: Select the closest value tolerance capacitor.

Setting the SLG59H1017V's Active Current Limit

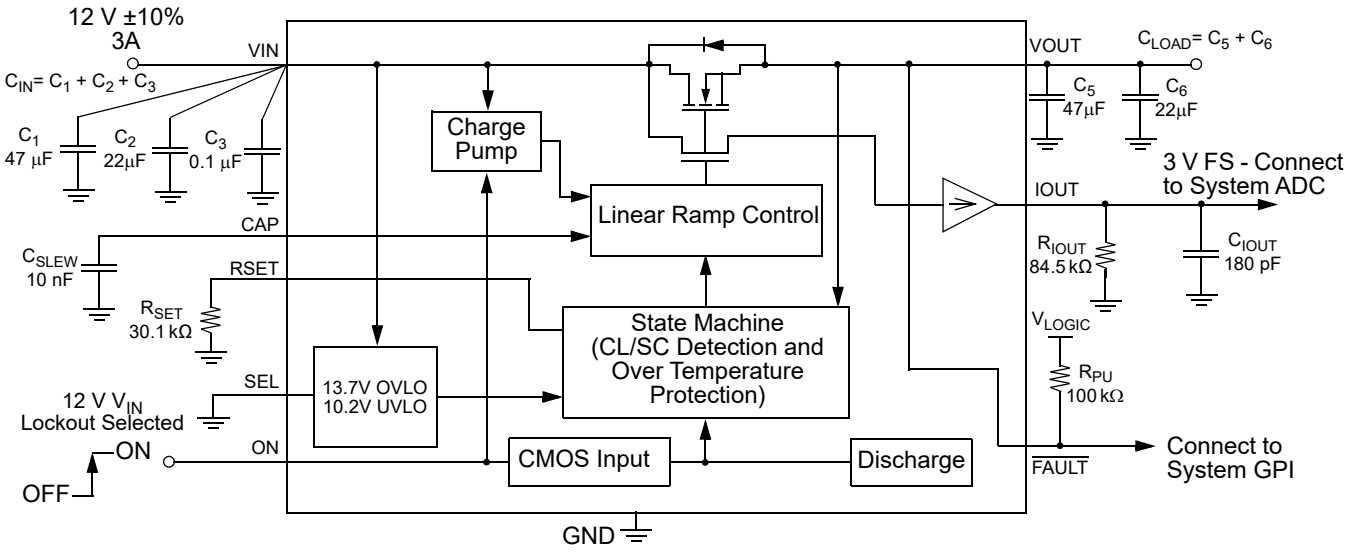
R_{SET} (kΩ)	Active Current Limit (A)[3]
91	1
45	2
30	3
20	4.5

Note 3: Active Current Limit accuracy is ±15% over voltage range and over temperature range.

Configuring the SLG59H1017V for 12 V V_{IN} Lockout Applications

To configure the SLG59H1017V for conditioned 12 V ±10% V_{IN} applications is simply a matter of connecting the SEL pin to GND as shown in Figure A. For other V_{IN} lockout window applications, please consult Dialog for additional information.

Figure A.



24 V V_{IN} and 12 V V_{IN} Lockout Window Thresholds

Shown in Figure B and Figure C are the two sets of V_{IN} overvoltage/undervoltage lockout windows – one for conditioned 24 V ±5% V_{IN} systems and the second for conditioned 12 V ±10% V_{IN} systems. To avoid lockout threshold collision with nominal operation, the SLG59H1017V's $V_{IN(OVLO)}$ min and $V_{IN(UVLO)}$ max thresholds were set 0.1 V correspondingly higher than the system's nominal V_{IN} max or lower than the system's V_{IN} min range.

Figure B.

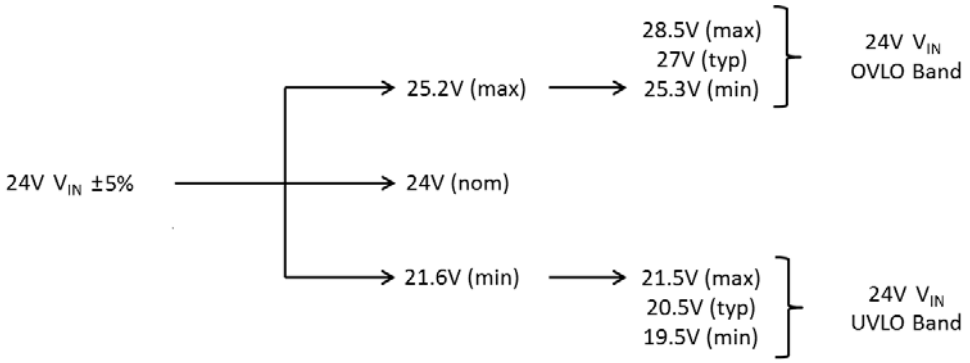
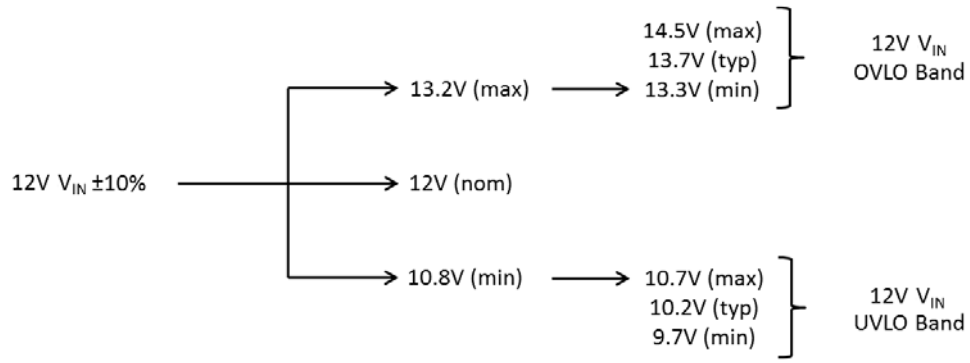


Figure C.



Power Dissipation

The junction temperature of the SLG59H1017V depends on different factors such as board layout, ambient temperature, and other environmental factors. The primary contributor to the increase in the junction temperature of the SLG59H1017V is the power dissipation of its power MOSFET. Its power dissipation and the junction temperature in nominal operating mode can be calculated using the following equations:

$$PD = RDS_{ON} \times I_{DS}^2$$

where:

PD = Power dissipation, in Watts (W)
RDS_{ON} = Power MOSFET ON resistance, in Ohms (Ω)
I_{DS} = Output current, in Amps (A)

and

$$T_J = PD \times \theta_{JA} + T_A$$

where:

T_J = Junction temperature, in Celsius degrees (°C)
θ_{JA} = Package thermal resistance, in Celsius degrees per Watt (°C/W)
T_A = Ambient temperature, in Celsius degrees (°C)

Power Dissipation (continued)

In current-limit mode, the SLG59H1017V's power dissipation can be calculated by taking into account the voltage drop across the power switch (V_{IN} - V_{OUT}) and the magnitude of the output current in current-limit mode (I_{ACL}):

$$PD = (V_{IN}-V_{OUT}) \times I_{ACL} \text{ or}$$

$$PD = (V_{IN} – (R_{LOAD} \times I_{ACL})) \times I_{ACL}$$

where:

PD = Power dissipation, in Watts (W)
V_{IN} = Input Voltage, in Volts (V)
R_{LOAD} = Load Resistance, in Ohms (Ω)
I_{ACL} = Output limited current, in Amps (A)
$V_{OUT} = R_{LOAD} \times I_{ACL}$

Layout Guidelines:

1. Since the VIN and VOUT pins dissipate most of the heat generated during high-load current operation, it is highly recommended to make power traces as short, direct, and wide as possible. A good practice is to make power traces with absolute minimum widths of 15 mils (0.381 mm) per Ampere. A representative layout, shown in Figure 26, illustrates proper techniques for heat to transfer as efficiently as possible out of the device;

2. To minimize the effects of parasitic trace inductance on normal operation, it is recommended to connect input C_{IN} and output C_{LOAD} low-ESR capacitors as close as possible to the SLG59H1017V's VIN and VOUT pins;

3. The GND pin should be connected to system analog or power ground plane.

4. 2 oz. copper is recommended for high current operation.

SLG59H1017V Evaluation Board:

A HFET1 Evaluation Board for SLG59H1017V is designed according to the statements above and is illustrated on Figure 26. Please note that evaluation board has D_Sense and S_Sense pads. They cannot carry high currents and dedicated only for RDS_{ON} evaluation.

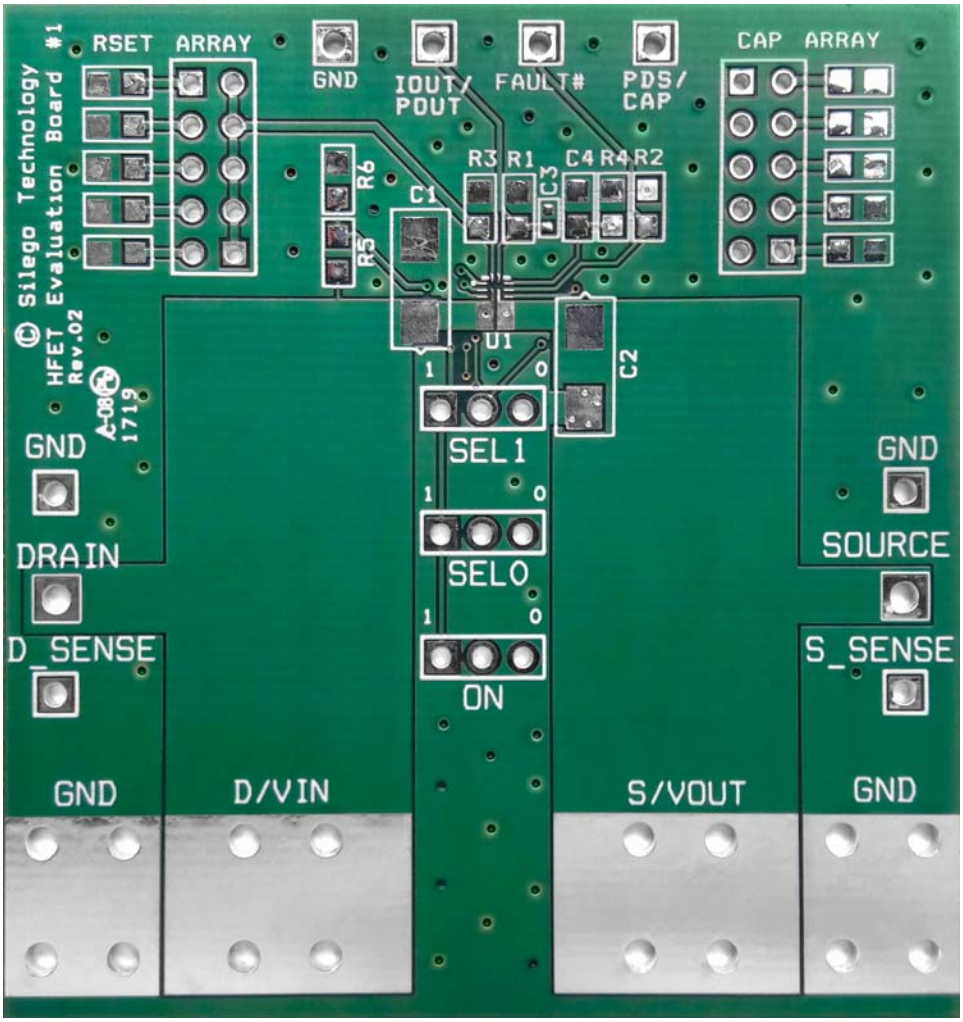


Figure 26. SLG59H1017V Evaluation Board

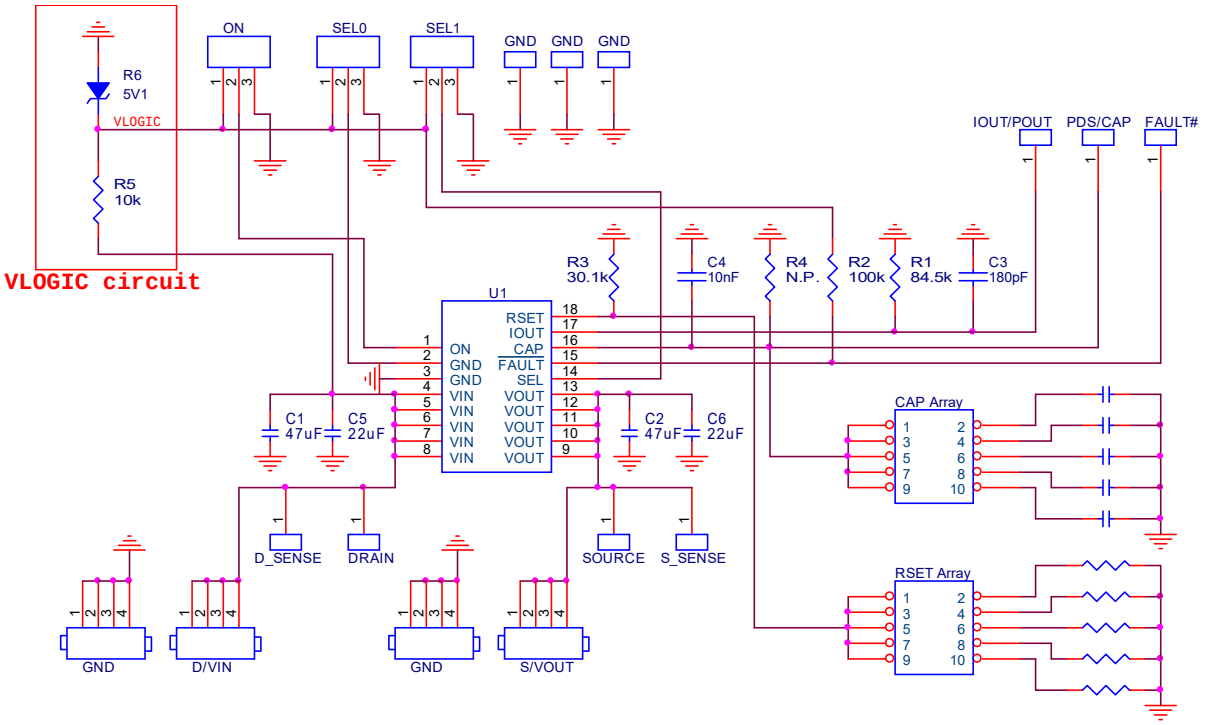


Figure 27. SLG59H1017V Evaluation Board Connection Circuit

Basic Test Setup and Connections

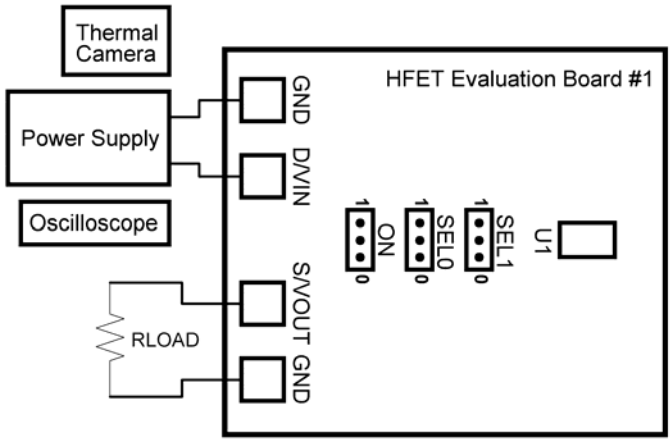
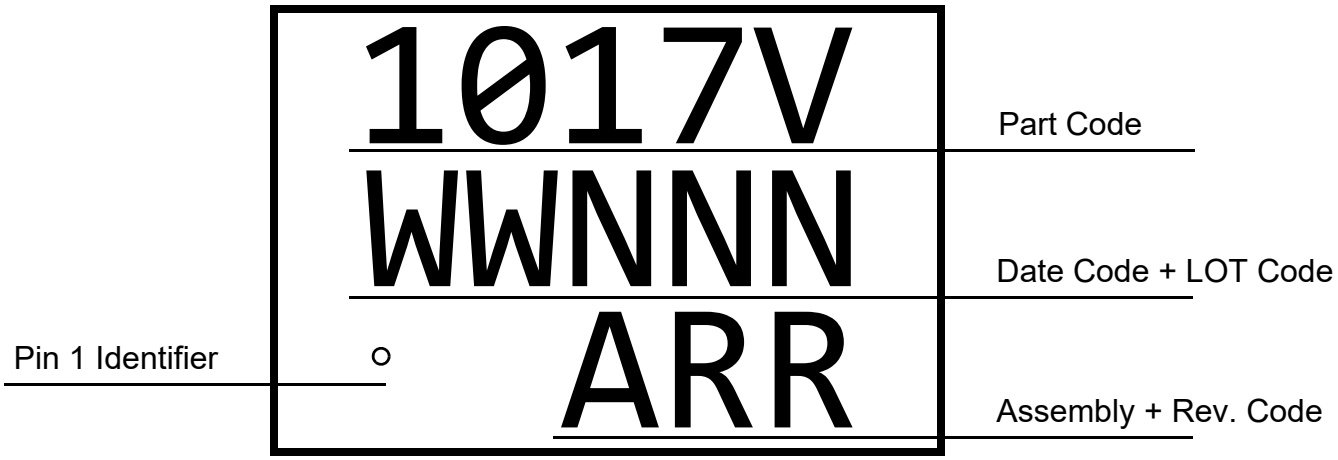


Figure 28. SLG59H1017V Evaluation Board Connection Circuit

EVB Configuration

1. Set SEL0 to GND;
2. Based on V_{IN} voltage, set SEL1 to GND or 5 V to configure OVLO;
3. Connect oscilloscope probes to D/VIN, S/VOUT, ON, etc.;
4. Turn on Power Supply and set V_{IN} to 12 V or 24 V;
5. Toggle the ON signal High or Low to observe SLG59H1017V operation.

Package Top Marking System Definition

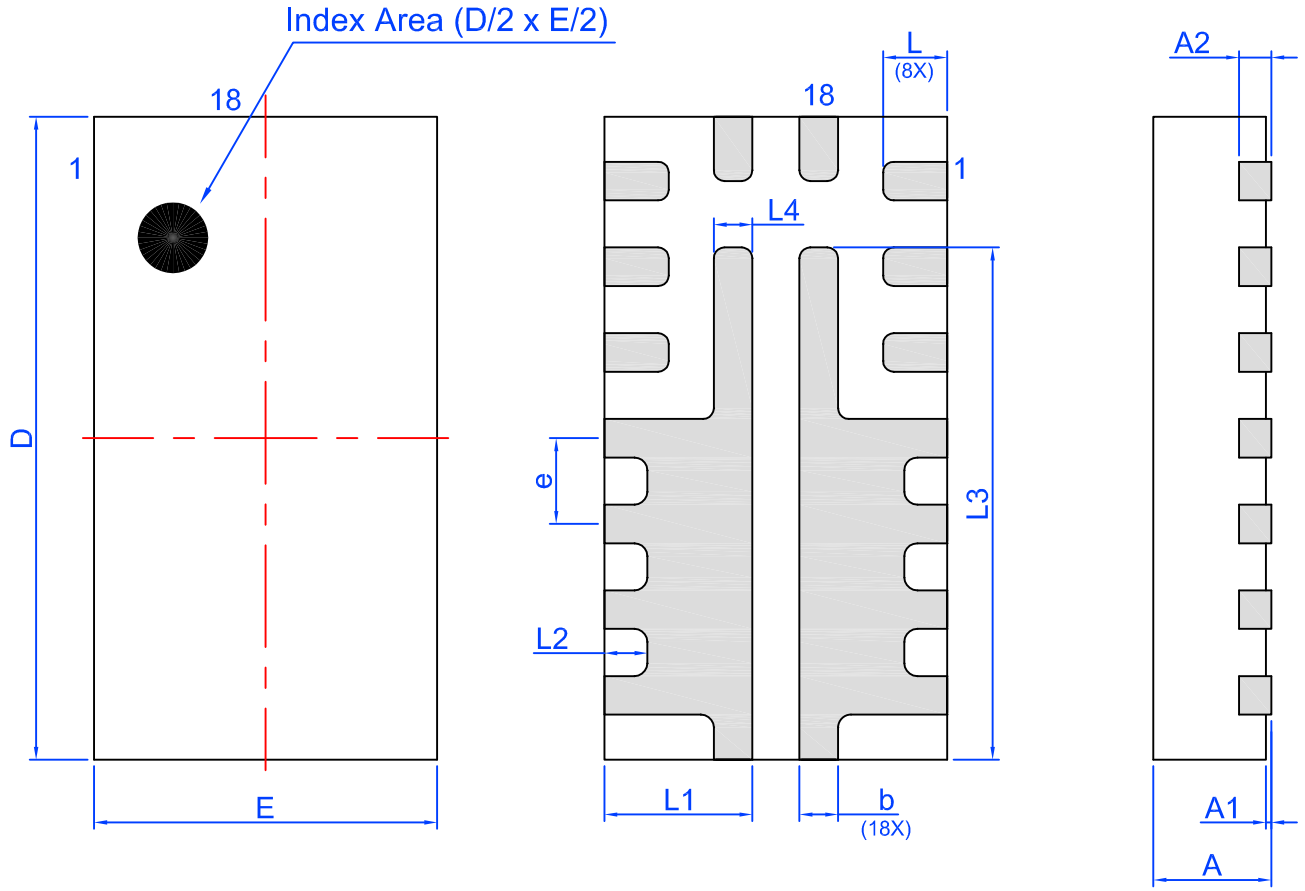


1017V - Part ID Field
WW - Date Code Field[1]
NNN - Lot Traceability Code Field[1]
A - Assembly Site Code Field [2]
RR - Part Revision Code Field[2]

Note 1: Each character in code field can be alphanumeric A-Z and 0-9
Note 2: Character in code field can be alphabetic A-Z

Package Drawing and Dimensions

18 Lead TQFN Package 1.6 x 3 mm (Fused Lead)
JEDEC MO-220, Variation WCEE

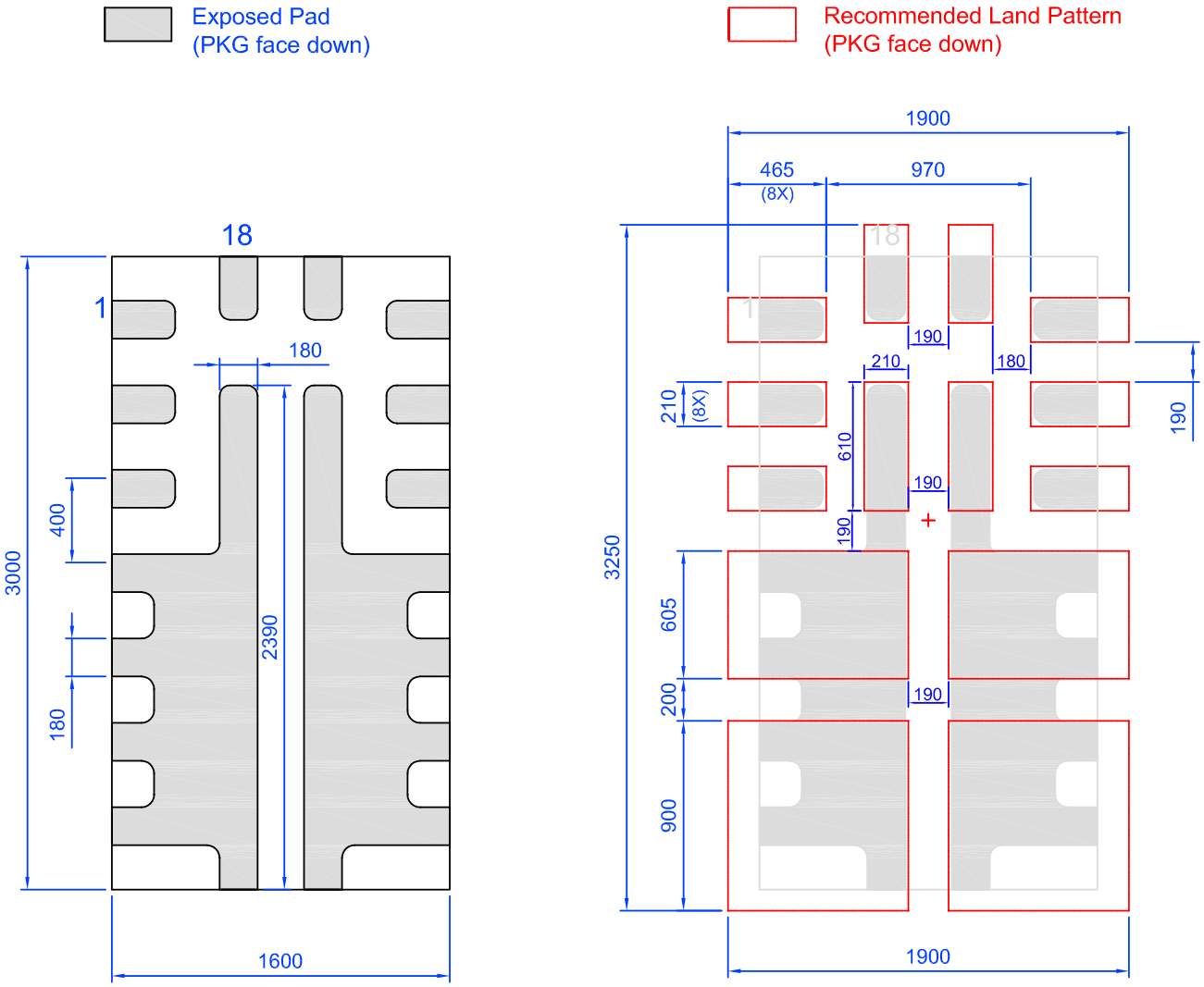


Top View BTM View Side View

Unit: mm

Symbol	Min	Nom.	Max	Symbol	Min	Nom.	Max
A	0.50	0.55	0.60	D	2.95	3.00	3.05
A1	0.005	-	0.05	E	1.55	1.60	1.65
A2	0.10	0.15	0.20	L	0.25	0.30	0.35
b	0.13	0.18	0.23	L1	0.64	0.69	0.74
e	0.40 BSC			L2	0.15	0.20	0.25
L3	2.34	2.39	2.44	L4	0.13	0.18	0.23

SLG59H1017V 18-pin STQFN PCB Landing Pattern



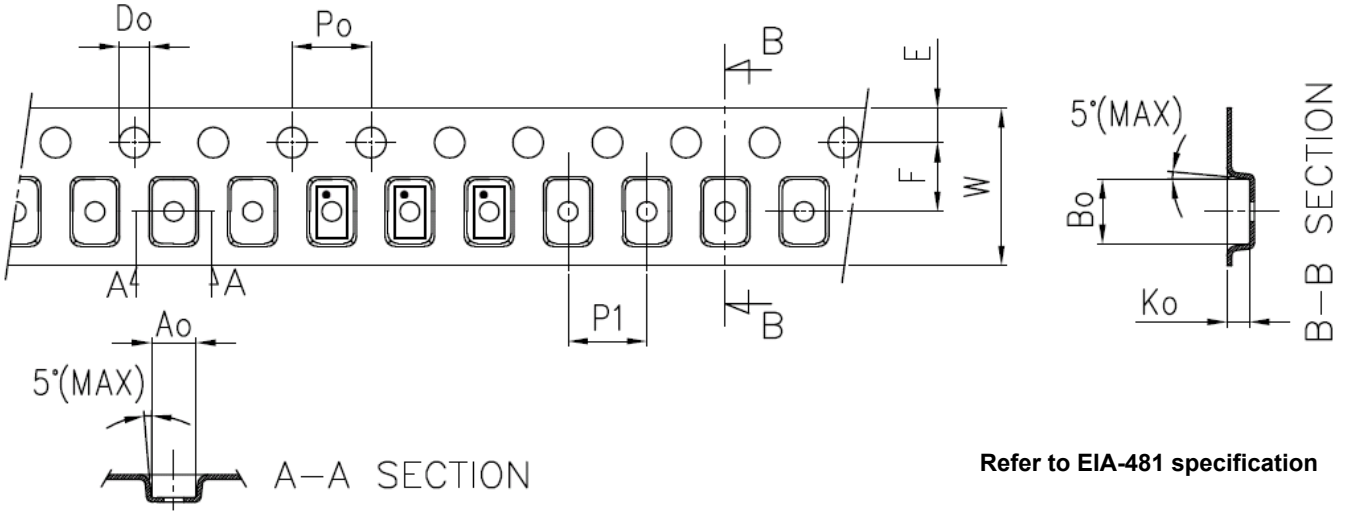
Note: All dimensions shown in micrometers (μm)

Tape and Reel Specifications

Package Type	# of Pins	Nominal Package Size [mm]	Max Units		Reel & Hub Size [mm]	Leader (min)		Trailer (min)		Tape Width [mm]	Part Pitch [mm]
			per Reel	per Box		Pockets	Length [mm]	Pockets	Length [mm]		
STQFN 18L 1.6x3mm 0.4P FC Green	18	1.6 x 3 x 0.55	3,000	3,000	178 / 60	100	400	100	400	8	4

Carrier Tape Drawing and Dimensions

Package Type	Pocket BTM Length	Pocket BTM Width	Pocket Depth	Index Hole Pitch	Pocket Pitch	Index Hole Diameter	Index Hole to Tape Edge	Index Hole to Pocket Center	Tape Width
	A0	B0	K0	P0	P1	D0	E	F	W
STQFN 18L 1.6x3mm 0.4P FC Green	1.78	3.18	0.76	4	4	1.5	1.75	3.5	8



Refer to EIA-481 specification

Recommended Reflow Soldering Profile

Please see IPC/JEDEC J-STD-020: latest revision for reflow profile based on package volume of 2.64 mm^3 (nominal). More information can be found at www.jedec.org.

Revision History

Date	Version	Change
10/17/2019	1.03	Updated Applications Info SOA Description Updated HFET Evaluation Board image
12/18/2018	1.02	Updated style and formatting Updated Charts Added Scopeshots Added Layout Guidelines Fixed typos
11/2/2017	1.01	Updated V_{IN} Max and $V_{IN(OVLO)}$ Min Fixed typos and formatting
5/13/2016	1.00	Production Release